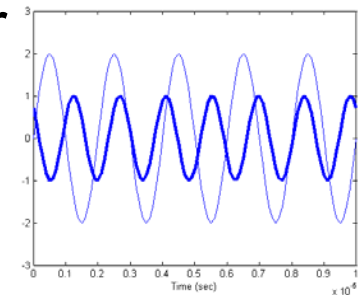


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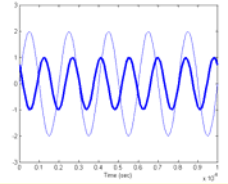
Signals and Systems

Spring 2006

Instructor: Dr. R. Michael Buehrer
Lecture #15: Applications of the Fourier
Transform – Part I: Frequency
Response and Ideal Filters

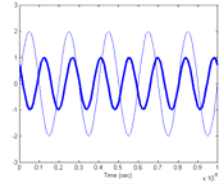


Overview



- We have discussed the Continuous Time Fourier Transform (CTFT) for a few lectures now.
- Today we examine several practical *applications* of the Fourier Transform
 - Frequency response and the Impulse response
 - Ideal filters
 - Lowpass, highpass, bandpass, bandstop
- What to read – Sections 6.1-6.3 in the text

Frequency Response



- Recall that an LTI system can be described by its *impulse response*
- Specifically

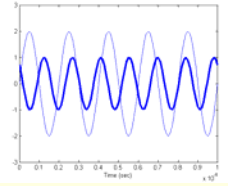
$$y(t) = \underbrace{x(t)}_{\text{input}} * \underbrace{h(t)}_{\text{impulse response}}$$

- In the frequency domain we have

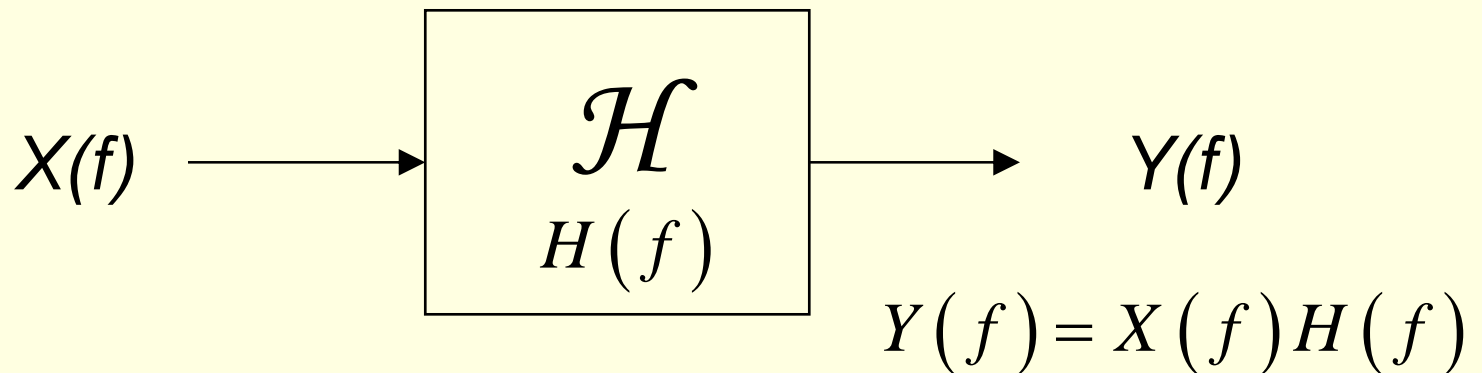
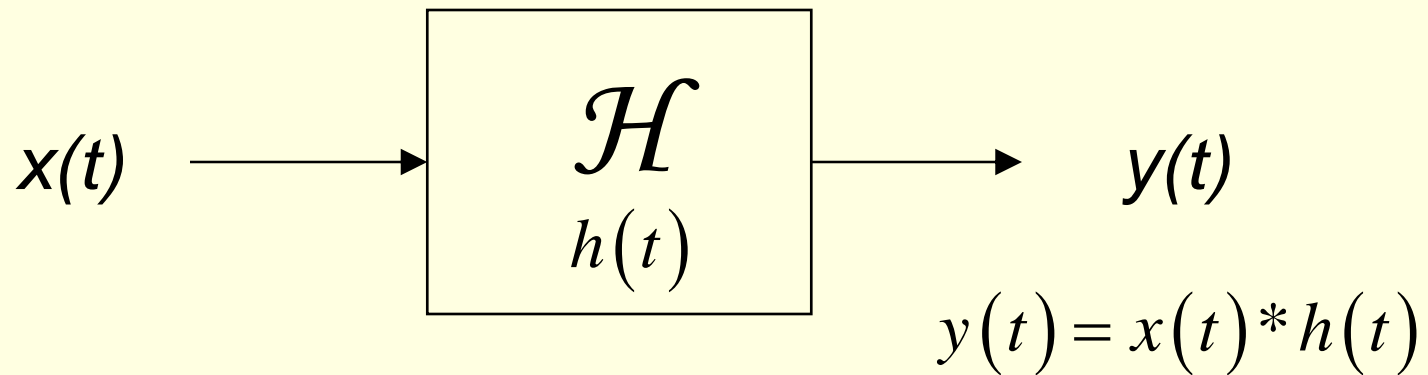
$$F\{y(t)\} = F\{x(t) * h(t)\}$$
$$Y(f) = X(f)H(f)$$

- The Fourier Transform of the impulse response, $H(f)$, is termed the *frequency response* of the system

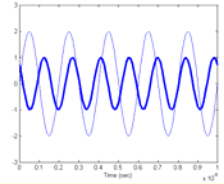
System Representation



- We can draw the system using the frequency domain or the time domain

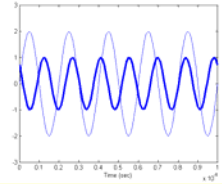


Distortion

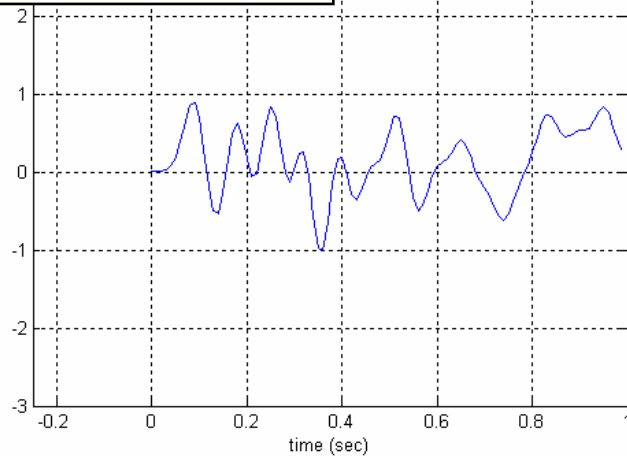


- *Distortion* is commonly construed to mean the shape of the signal has changed
- Amplifying the signal by a constant gain or delaying the signal is not distortion
- Examples of distortion
 - Non-constant (time-selective) gain
 - Frequency-selective gain
 - Clipping

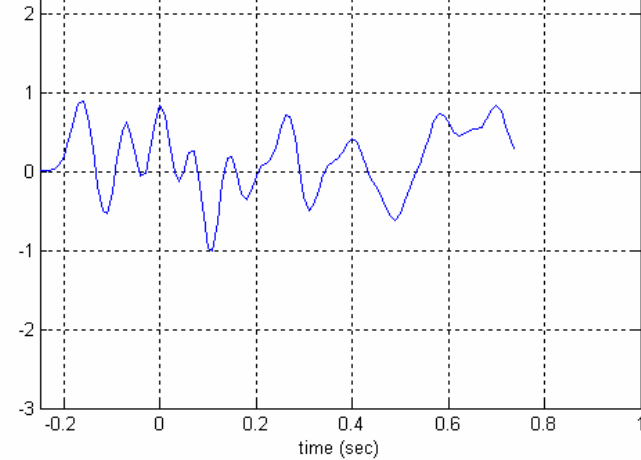
Changes that are not distortion



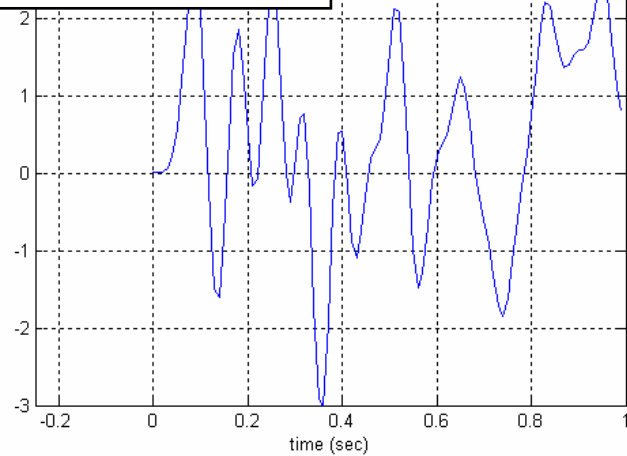
Original Signal



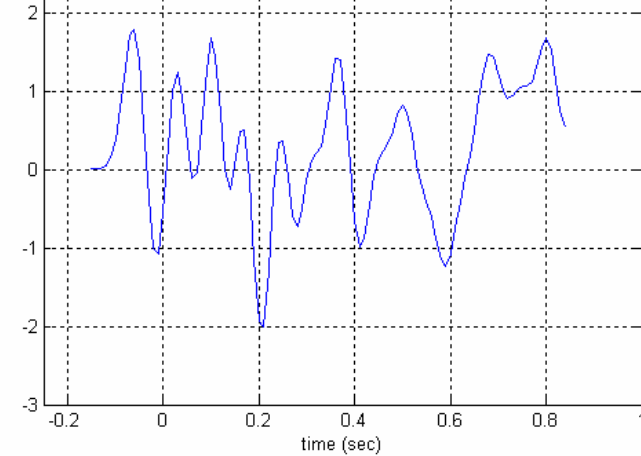
Time shift



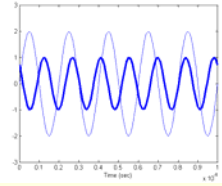
Constant Gain



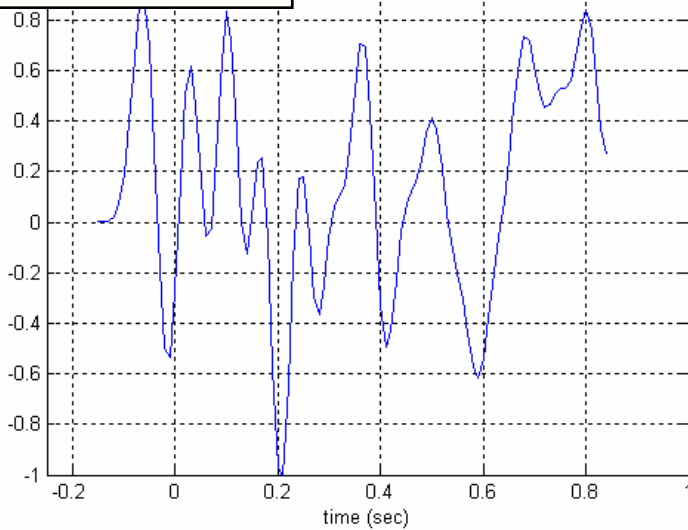
Constant Gain and Time Shift



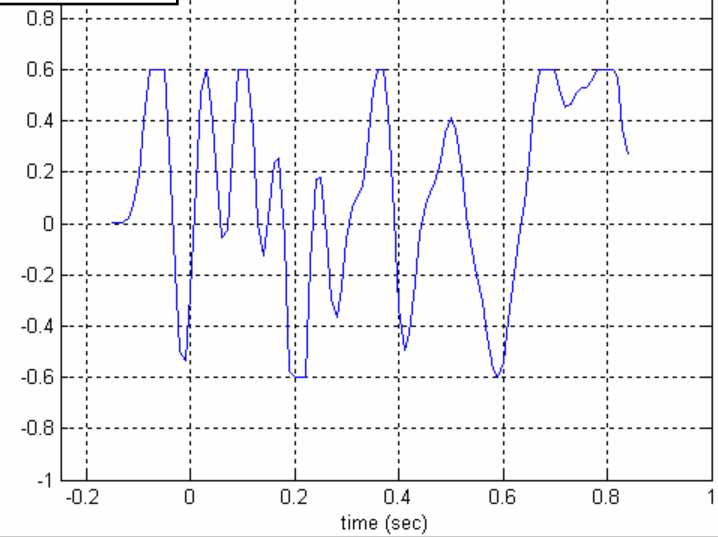
Examples of Distortion



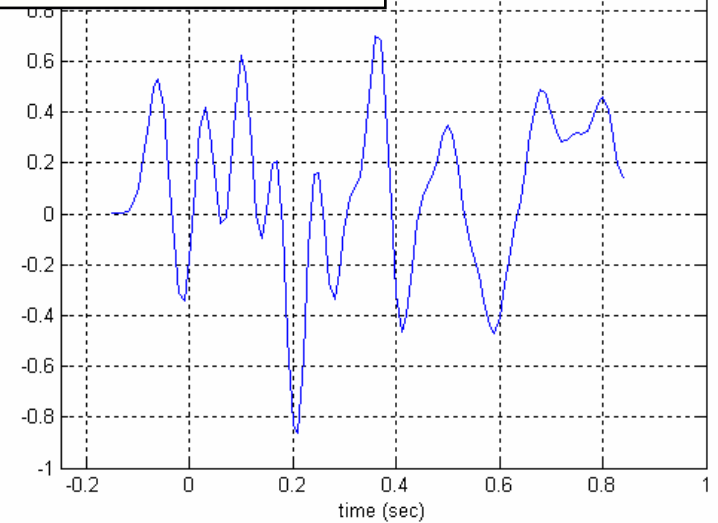
Original Signal



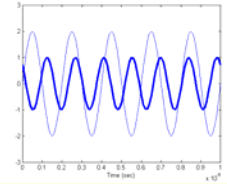
Clipping



Time-varying gain

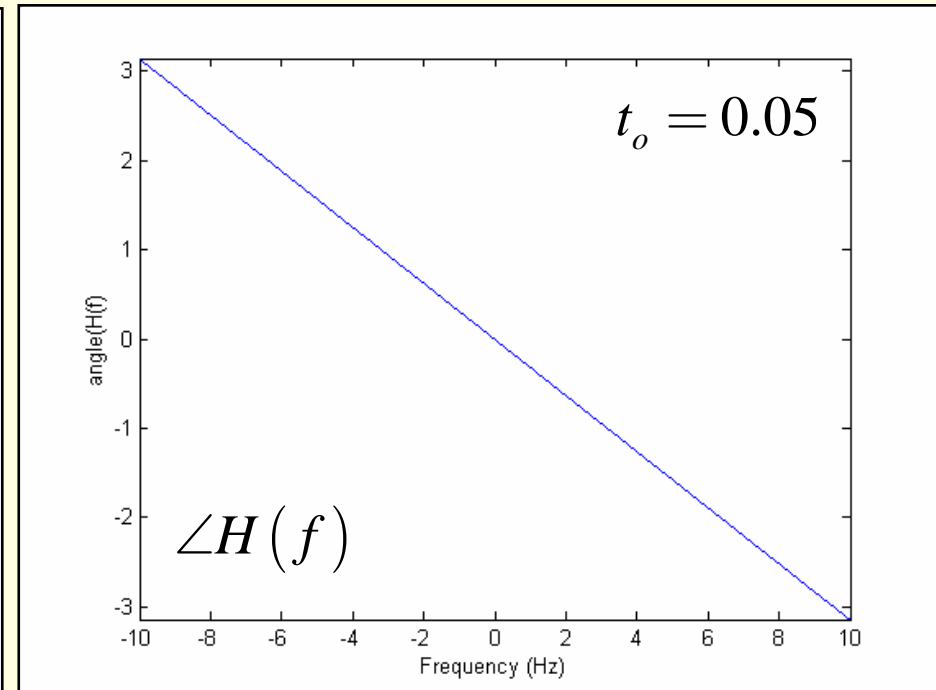
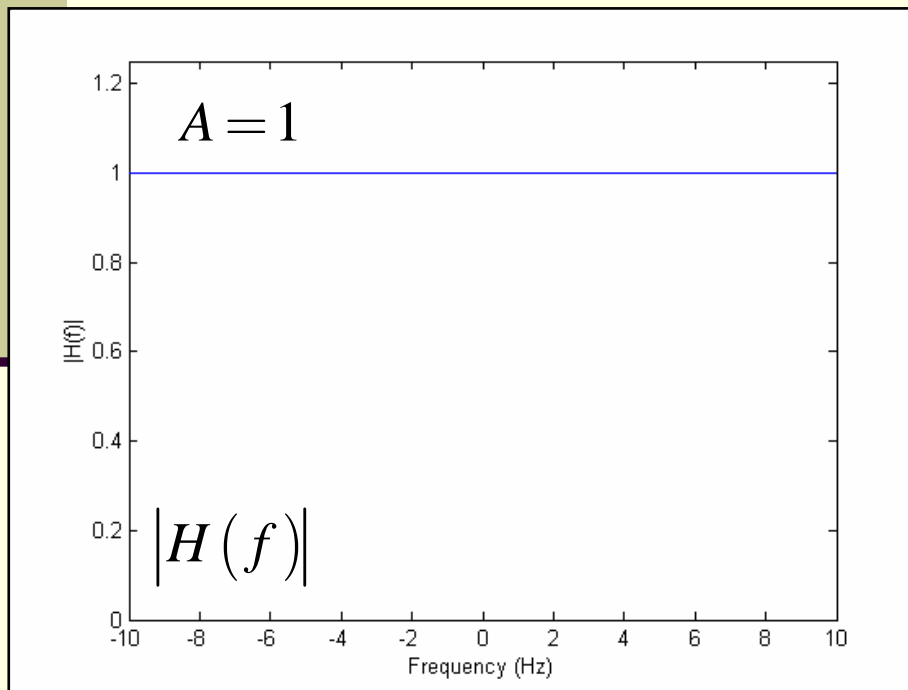


Distortionless System

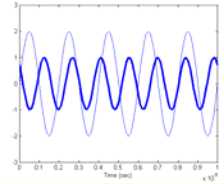


- A distortionless system passes all frequencies with equal gain and has a linear phase response (i.e., simple time shift)

$$H(f) = Ae^{-j2\pi ft_o}$$



Distortionless System



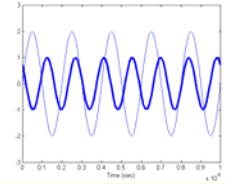
- The distortionless system was defined in the frequency domain
- What does this mean in the time domain?

$$H(f) = Ae^{-j2\pi ft_o}$$

- The impulse response of the distortionless system is found by taking the inverse FFT

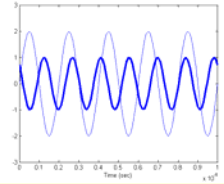
$$\begin{aligned}h(t) &= F^{-1} \{ H(f) \} \\ &= F^{-1} \{ Ae^{-j2\pi ft_o} \} \\ &= A\delta(t - t_o)\end{aligned}$$

Filters



- A filter is a system which passes certain frequencies and rejects other frequencies
- Types of filters
 - Low pass filter
 - High pass filter
 - Bandpass filter
 - Bandstop filter
- Ideal filter
 - An ideal filter is one which perfectly passes frequencies in a certain range (termed the *pass band*) and perfectly rejects frequencies in another range termed the *stop band*
 - An ideal filter doesn't *distort* the signal in the pass band

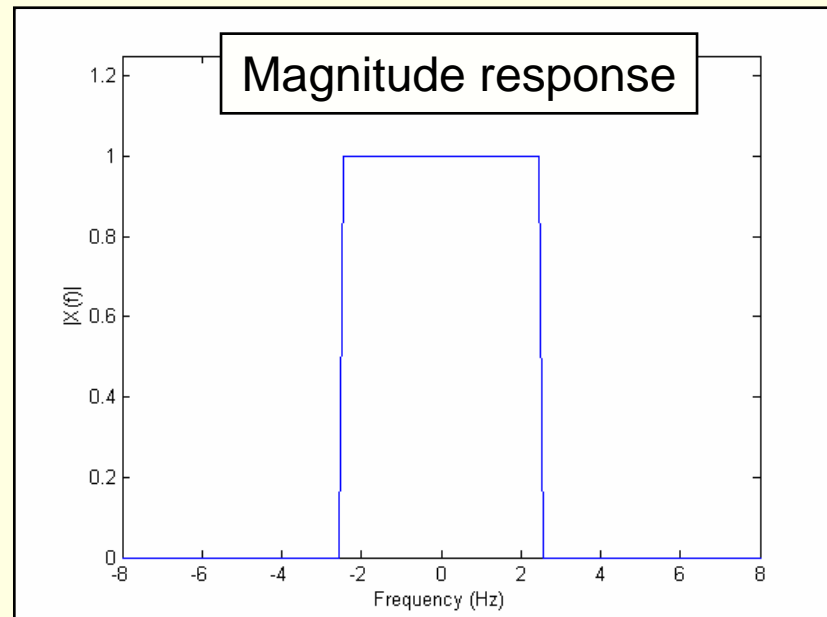
The Ideal Lowpass Filter



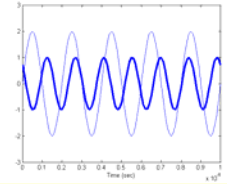
- The ideal lowpass filter has a frequency response

$$H(f) = \begin{cases} Ae^{-j2\pi ft_o} & |f| \leq f_m \\ 0 & |f| > f_m \end{cases}$$
$$= \underbrace{A \operatorname{rect}\left(\frac{f}{2f_m}\right)}_{\text{magnitude response}} \underbrace{e^{-j2\pi ft_o}}_{\text{time delay}}$$

Example – $f_m = 2.5\text{Hz}$

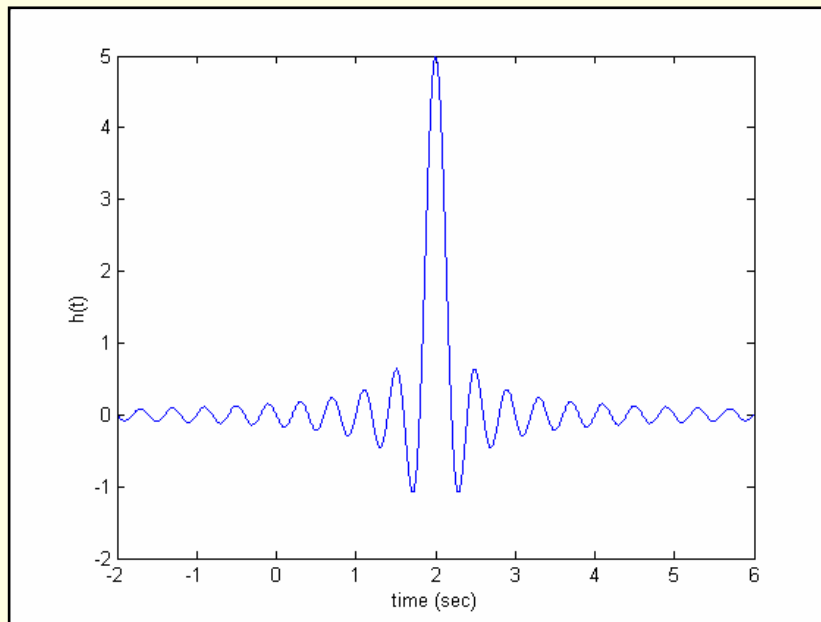


Impulse response of the Ideal LPF



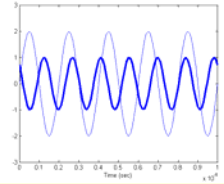
- The impulse response of the ideal LPF can be found by taking the inverse Fourier Transform of the ideal LPF frequency response

$$\begin{aligned}h(t) &= F^{-1}\{H(f)\} \\ &= F^{-1}\left\{A \operatorname{rect}\left(\frac{f}{2f_m}\right)e^{-j2\pi ft_o}\right\} \\ &= 2Af_m \operatorname{sinc}(2f_m(t-t_o))\end{aligned}$$



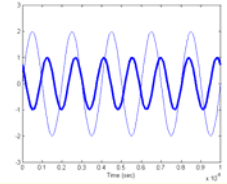
- Example: $f_m = 5$
- $t_o = 2$
- Is there any problem with this impulse response?

Causality

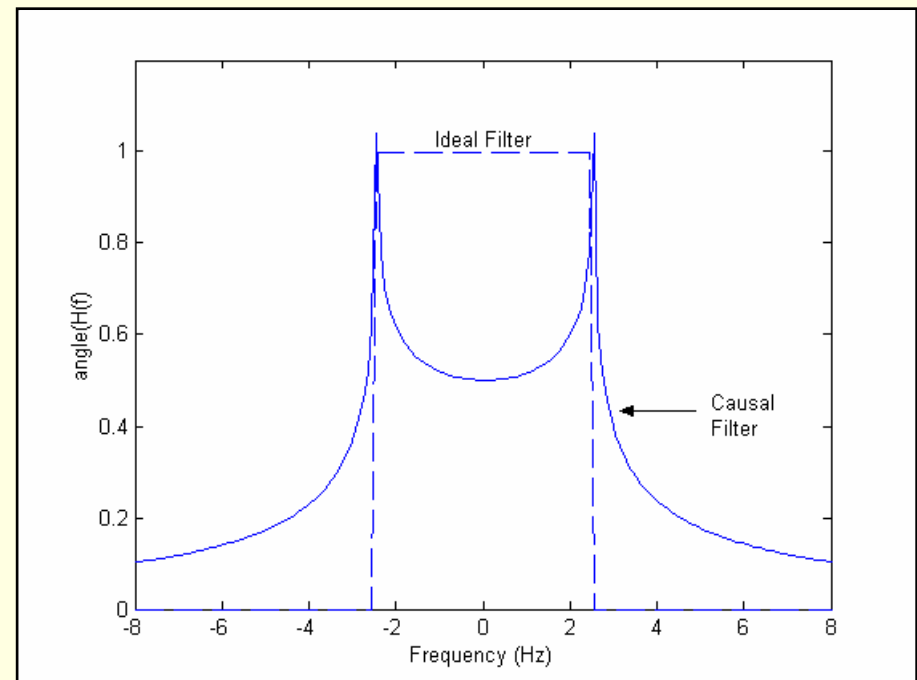
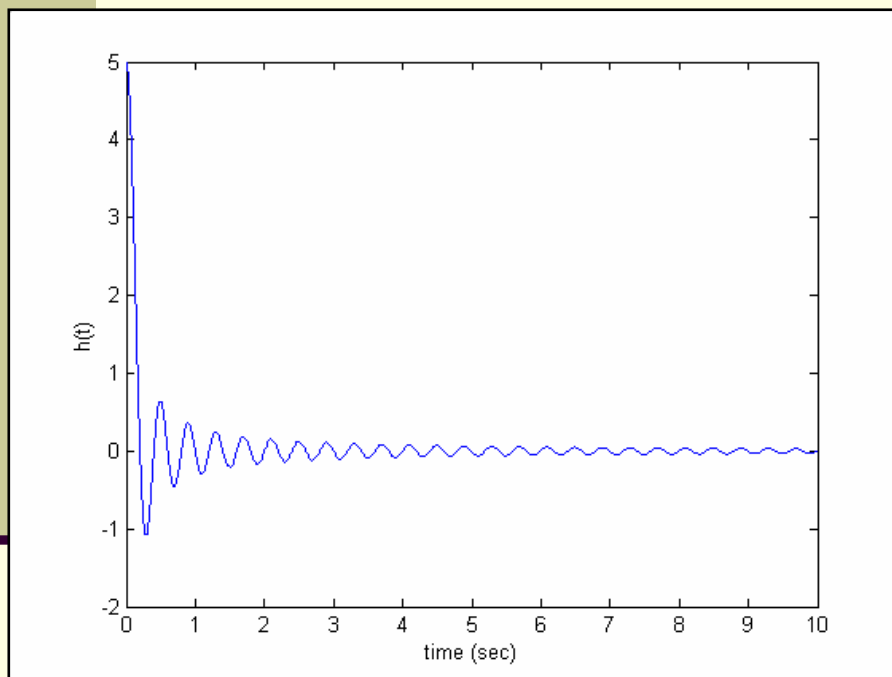


- A causal system is one that does not have a response prior to the time of the applied input
- The impulse response is the response of a system to an impulse applied at time $t=0$.
- A system whose impulse response is nonzero for $t < 0$ is thus *non-causal*.
- The ideal lowpass filter is thus non-causal and for that reason is not physically realizable.
- We can attempt to approximate the ideal LPF by introducing a delay into our system

Causal LPF

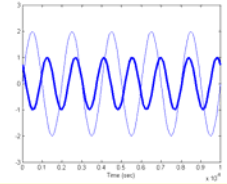


- We can make the filter causal by simply truncating the impulse response before $t=0$.

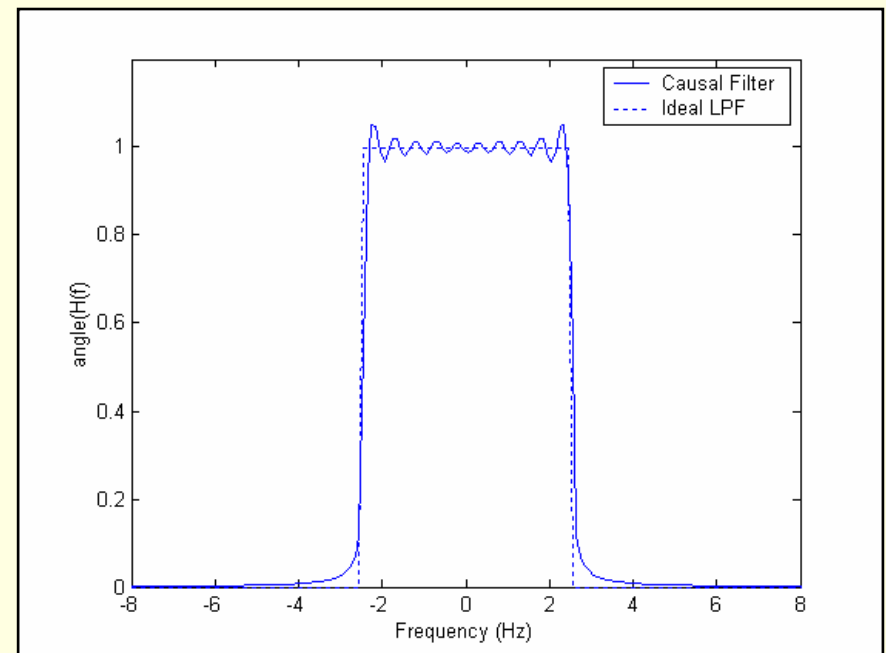
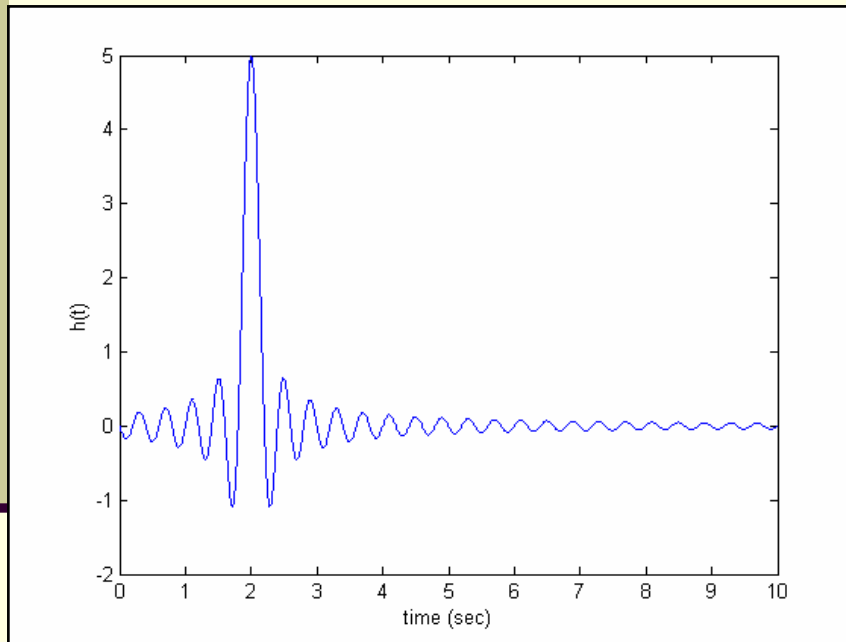


- The resulting filter is far from ideal.

Causal LPF

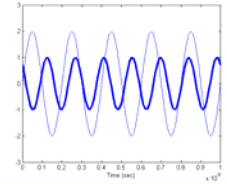


- A second option is to delay the impulse response and truncate it.
- Delay = 2

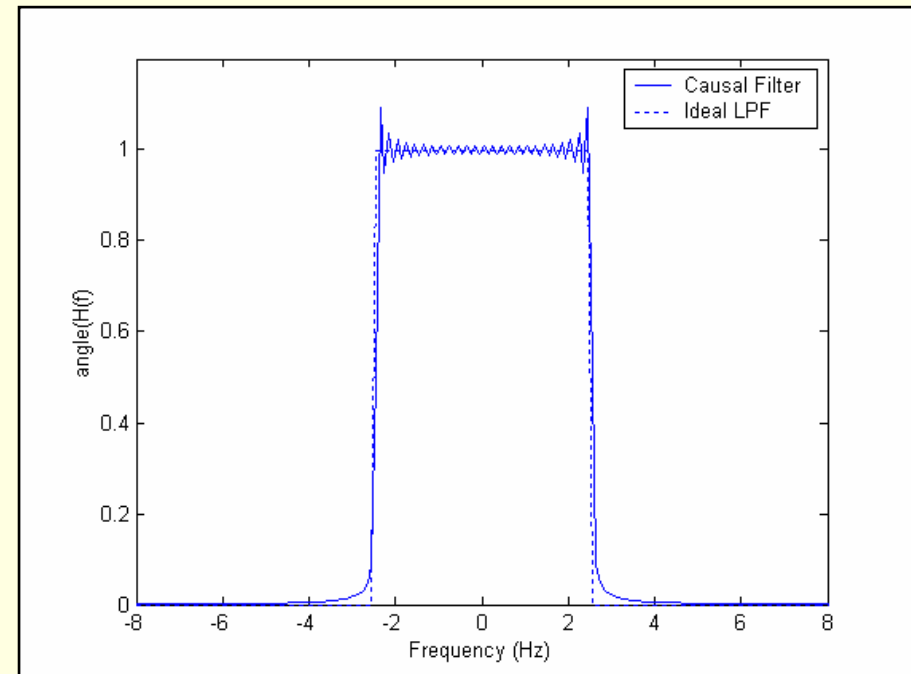
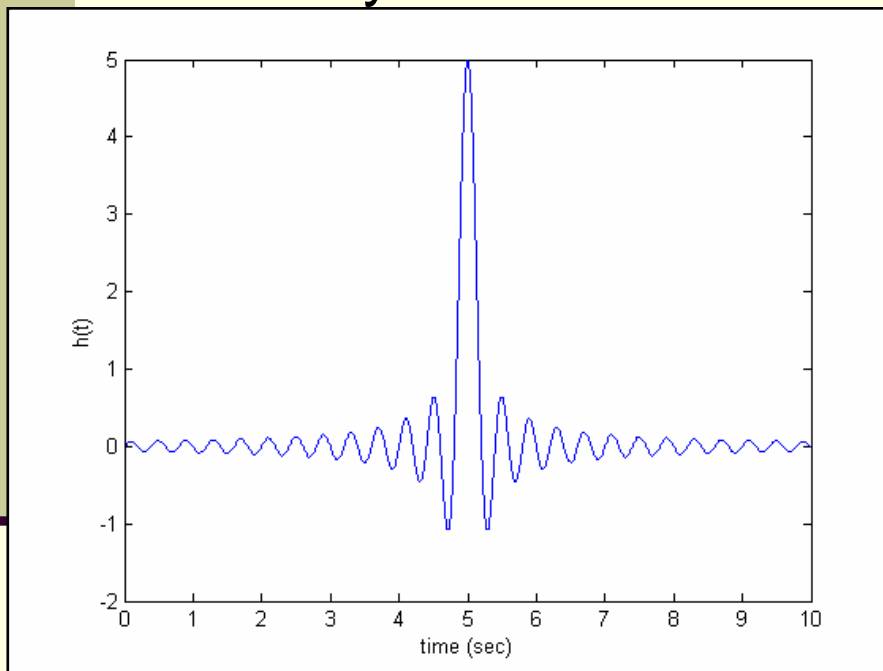


- This makes the system closer to ideal, but requires a delay which some applications may not tolerate

Causal LPF

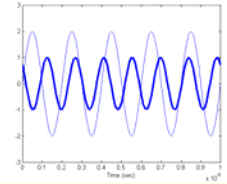


- A second option is to delay the impulse response and truncate it.
- Delay = 5

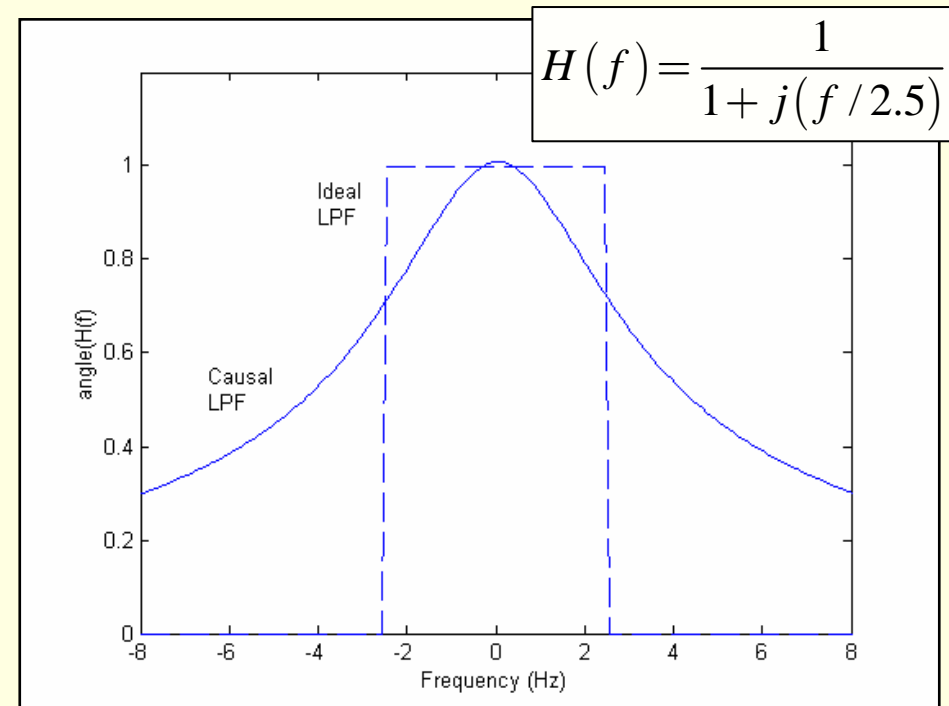
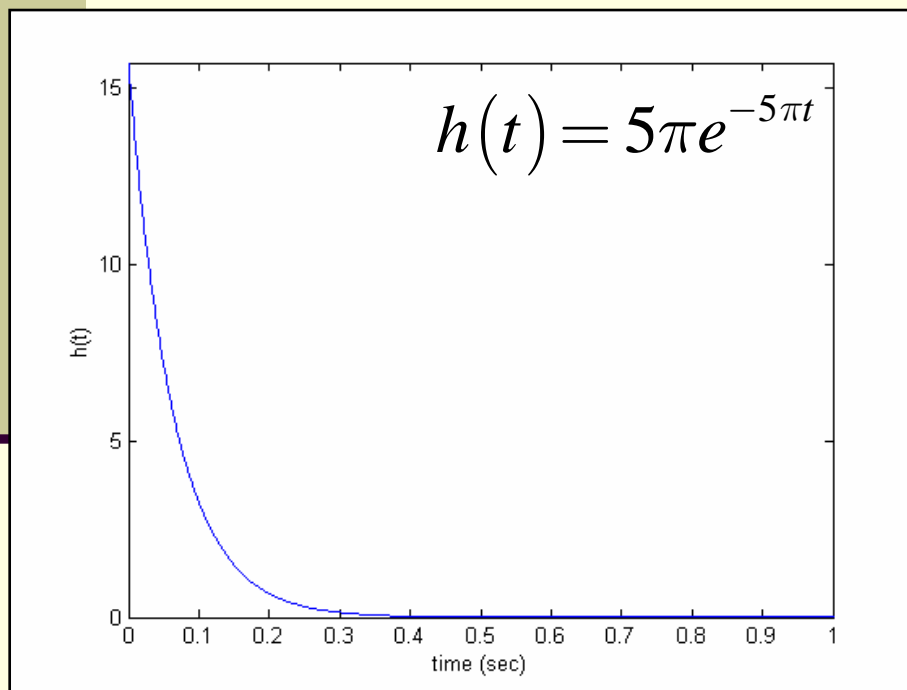


- Larger delay leads to better approximation of the ideal LPF

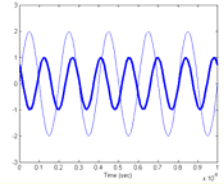
Causal LPF



- RC filter
- Causal, exponential impulse response
- Non-ideal LPF



Ideal High Pass Filter



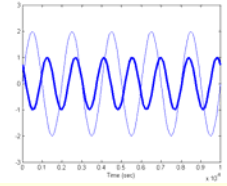
- The ideal HPF has a frequency response

$$H(f) = \begin{cases} 0 & |f| \leq f_m \\ Ae^{-j2\pi ft_o} & |f| > f_m \end{cases}$$
$$= \underbrace{\left[A - A \operatorname{rect}\left(\frac{f}{2f_m}\right) \right]}_{\substack{\text{magnitude} \\ \text{response}}} \underbrace{e^{-j2\pi ft_o}}_{\substack{\text{time} \\ \text{delay}}}$$

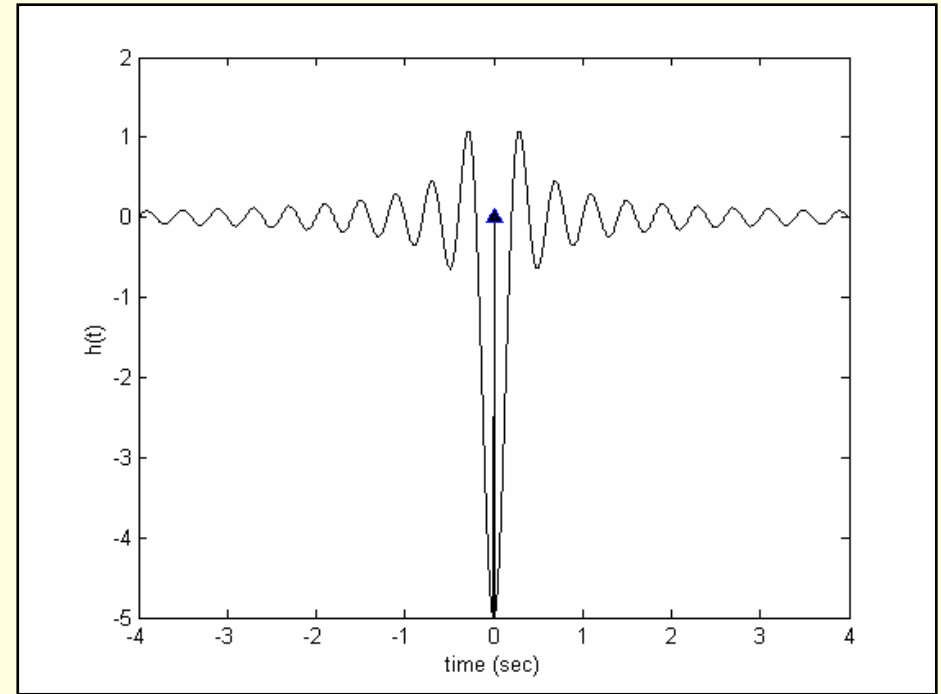
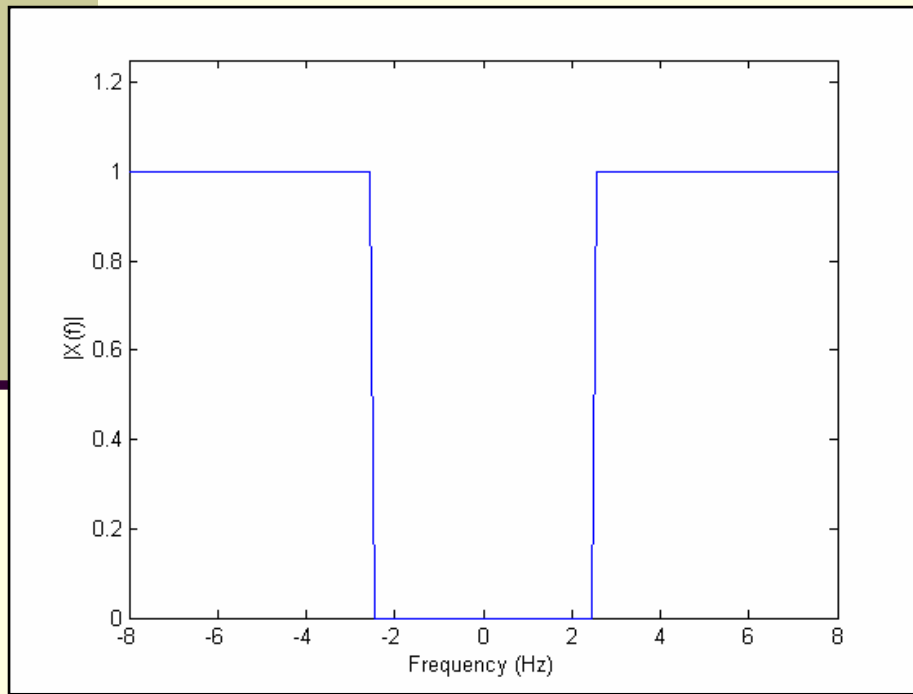
- The corresponding impulse response is

$$h(t) = F^{-1}\{H(f)\}$$
$$= F^{-1}\left\{ \left[A - A \operatorname{rect}\left(\frac{f}{2f_m}\right) \right] e^{-j2\pi ft_o} \right\}$$
$$= A\delta(f) - 2Af_m \operatorname{sinc}(2f_m(t - t_o))$$

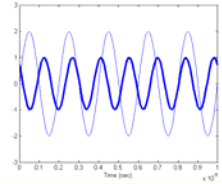
Ideal High Pass Filter



- The ideal HPF is non-causal, just like the ideal lowpass filter



Ideal Bandpass Filter



- The ideal bandpass filter can be written as

$$\begin{aligned}
 H(f) &= \begin{cases} A e^{-j2\pi f t_o} & f_L \leq |f| \leq f_H \\ 0 & \text{else} \end{cases} \\
 &= A \underbrace{\left[\text{rect}\left(\frac{f - f_o}{\Delta f}\right) + \text{rect}\left(\frac{f + f_o}{\Delta f}\right) \right]}_{\text{magnitude response}} \underbrace{e^{-j2\pi f t_o}}_{\text{time delay}}
 \end{aligned}$$

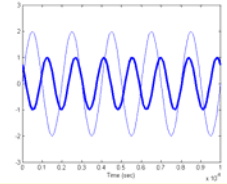
$$h(t) = F^{-1} \{ H(f) \}$$

$$= F^{-1} \left\{ A \left[\text{rect}\left(\frac{f - f_o}{\Delta f}\right) + \text{rect}\left(\frac{f + f_o}{\Delta f}\right) \right] e^{-j2\pi f t_o} \right\}$$

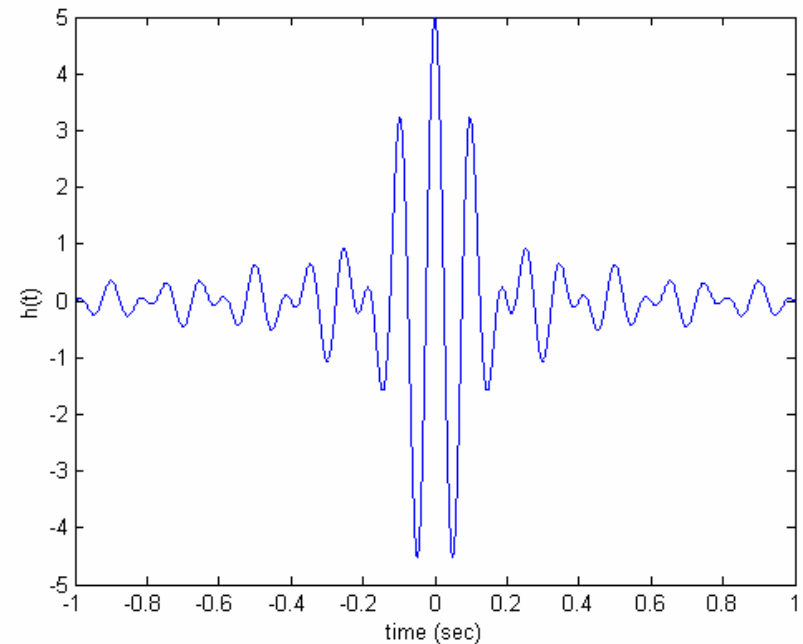
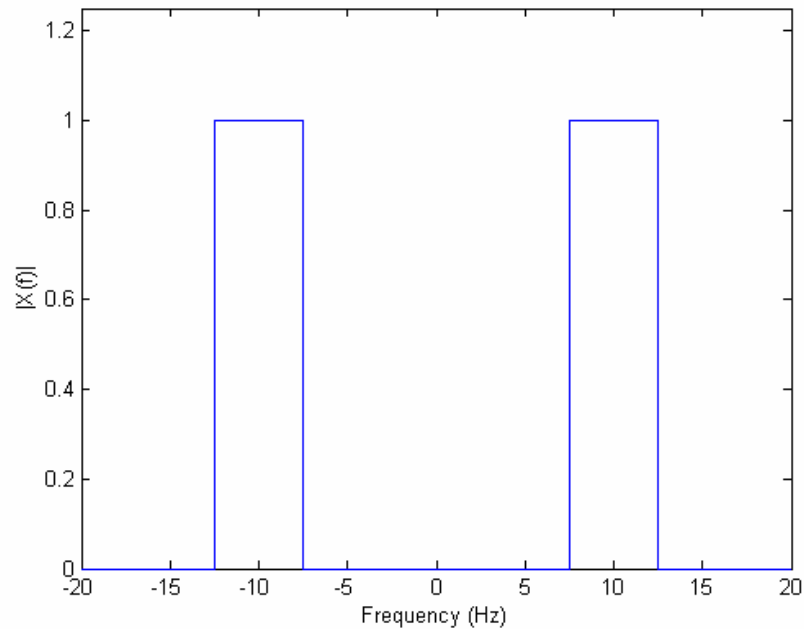
$$= 2A\Delta f \text{sinc}(\Delta f_m (t - t_o)) \cos(2\pi f_o (t - t_o))$$

f_L = lower frequency limit
 f_H = upper frequency limit
 $\Delta f = f_H - f_L$
 $f_o = (f_H + f_L)/2$

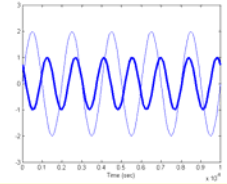
Ideal Bandpass Filter



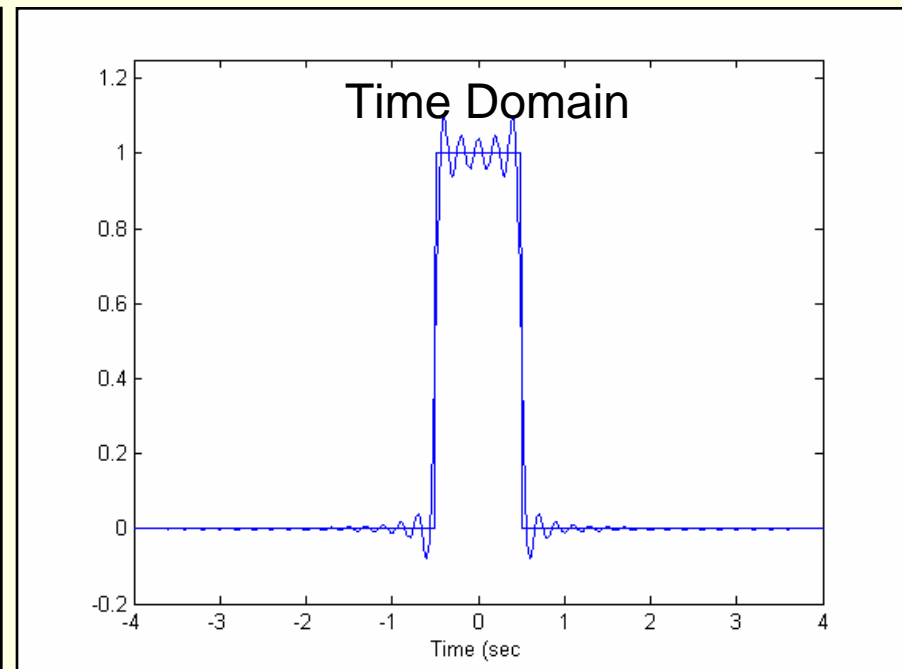
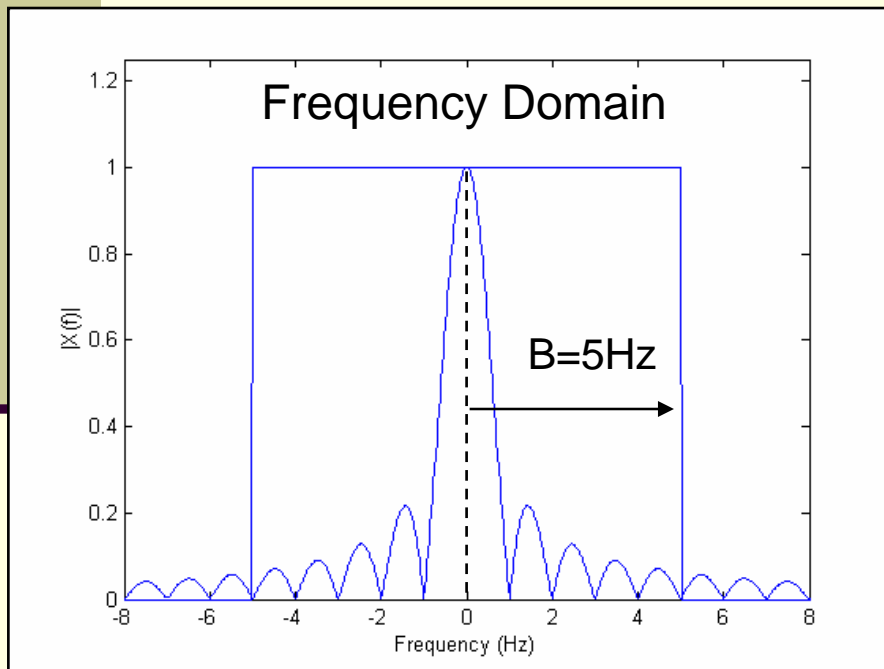
- Ideal BPF is also non-causal
- Realistic bandpass filters will be causal and not have a perfect frequency response



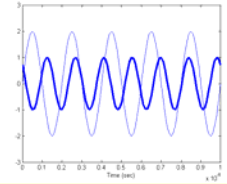
Example



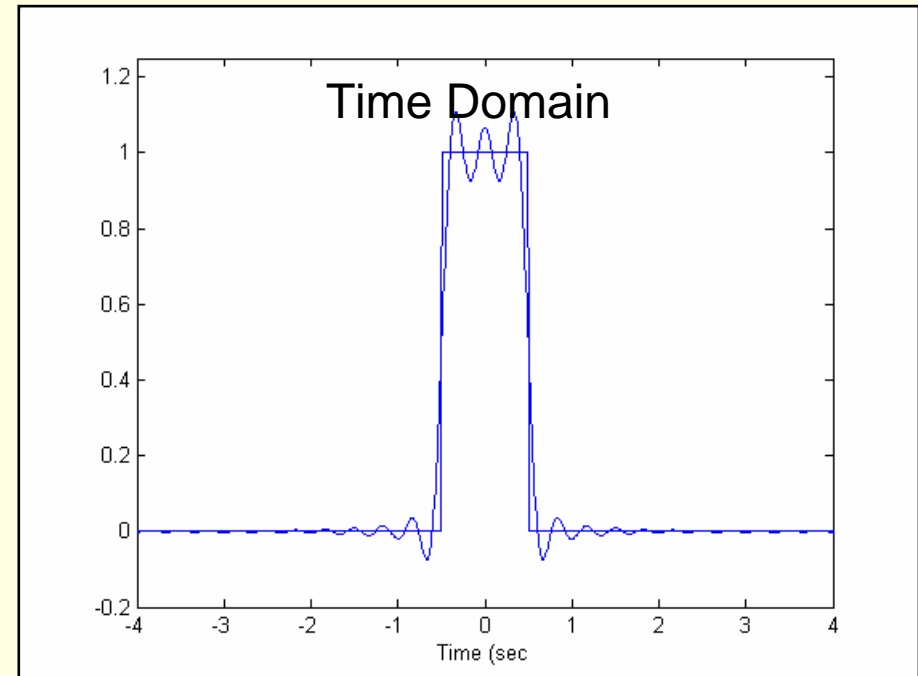
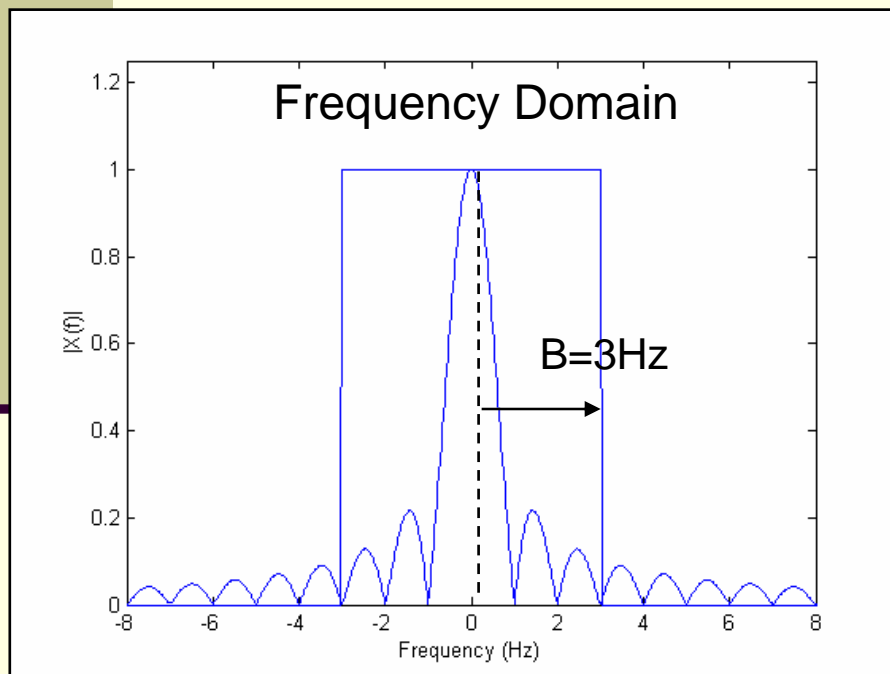
- Consider a time-domain square pulse of width 1 second which is passed through a filter with a bandwidth of 5Hz
- The bandwidth restriction does not cause a substantial change in the pulse shape



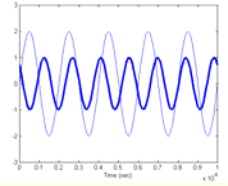
Example – cont.



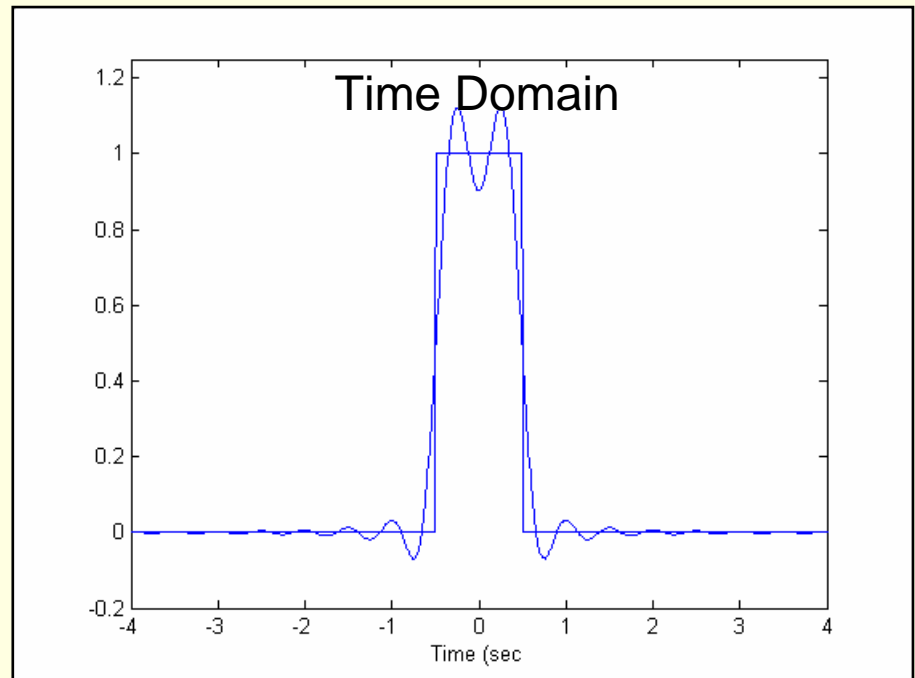
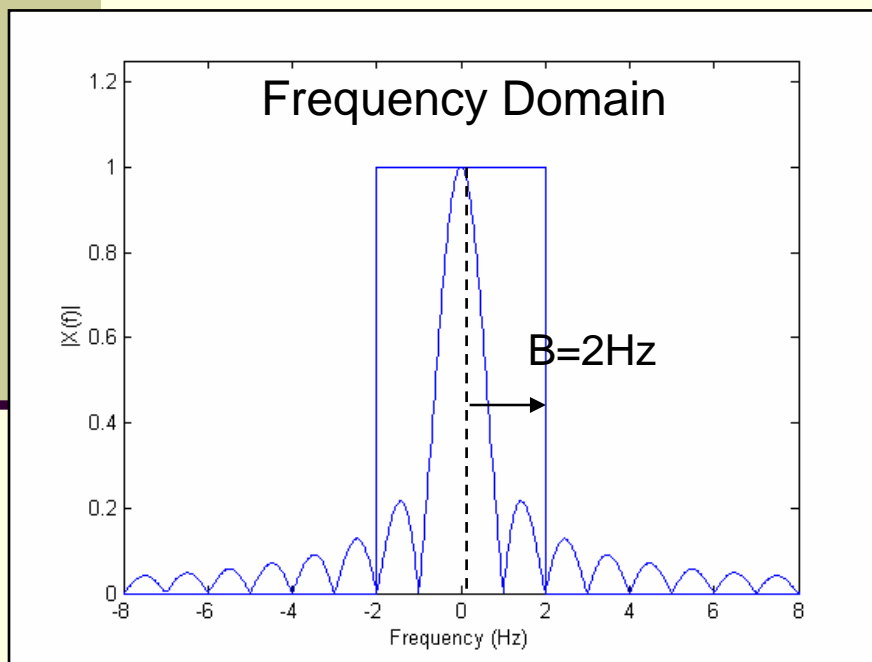
- Now consider a filter with a bandwidth of 3Hz
- The bandwidth restriction still does not cause a substantial change in the pulse shape
- Ringing occurs near sharp transitions



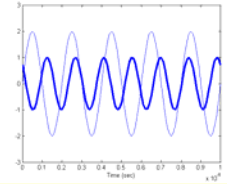
Example – cont.



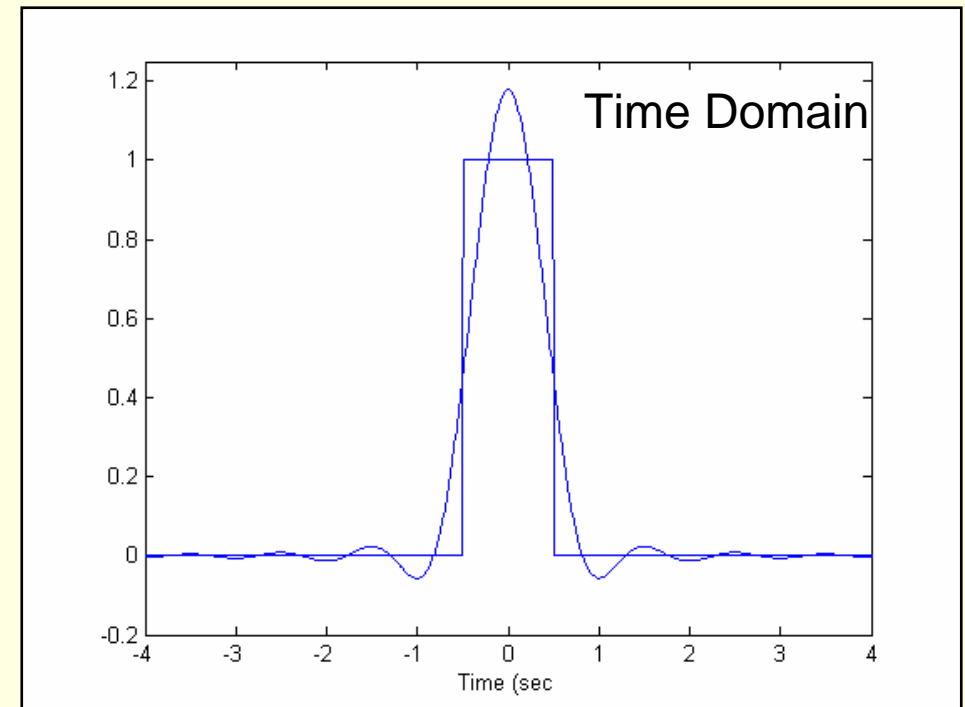
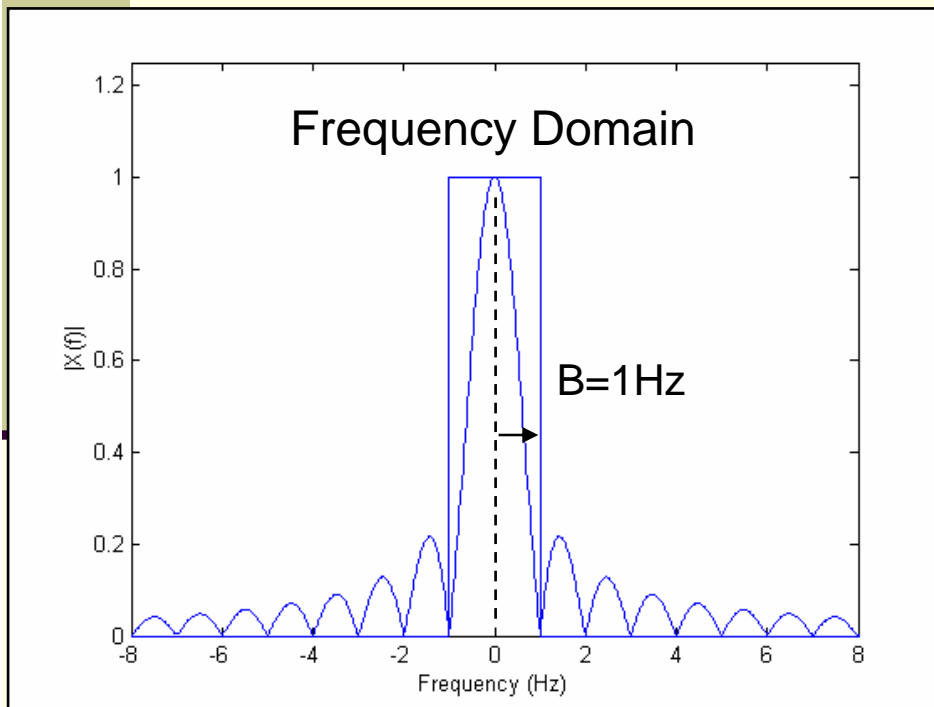
- Now consider a filter with a bandwidth of 2Hz
- The bandwidth restriction now begins to cause a more substantial change in the pulse shape
- Sharp transitions cannot occur



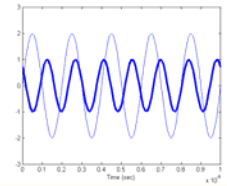
Example – cont.



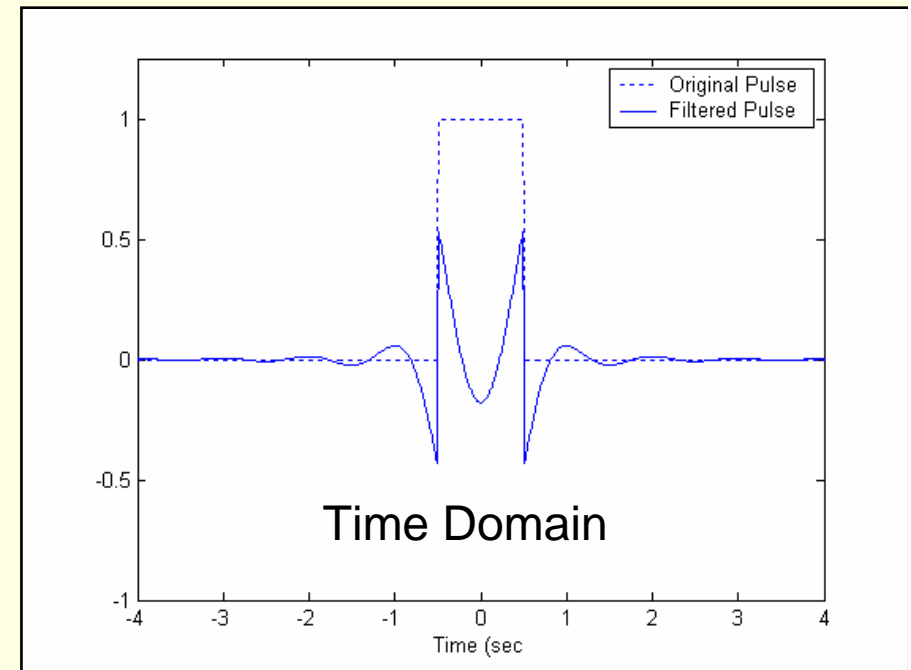
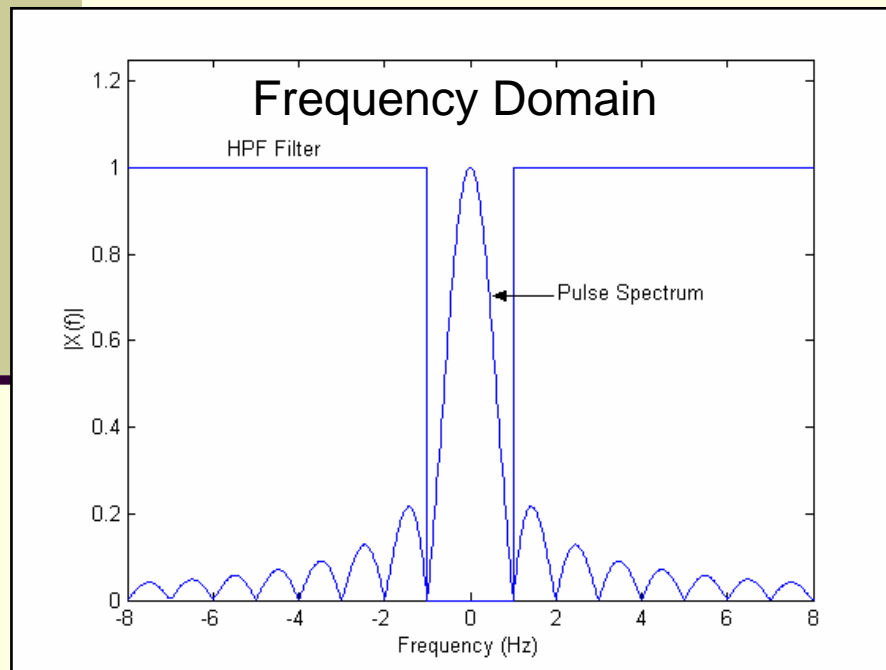
- Now consider a filter with a bandwidth of 1Hz
- The bandwidth restriction now changes the shape considerably



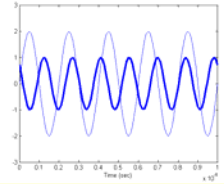
Example 2



- Consider a time-domain square pulse of width 1 second which is passed through a HP filter with a bandstop bandwidth of 1Hz
- The loss of the low-frequency energy causes a large change in the pulse. Note that the pulse now has zero average value.

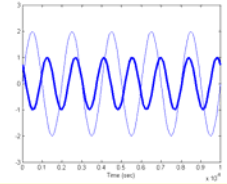


Example 3 - Noise

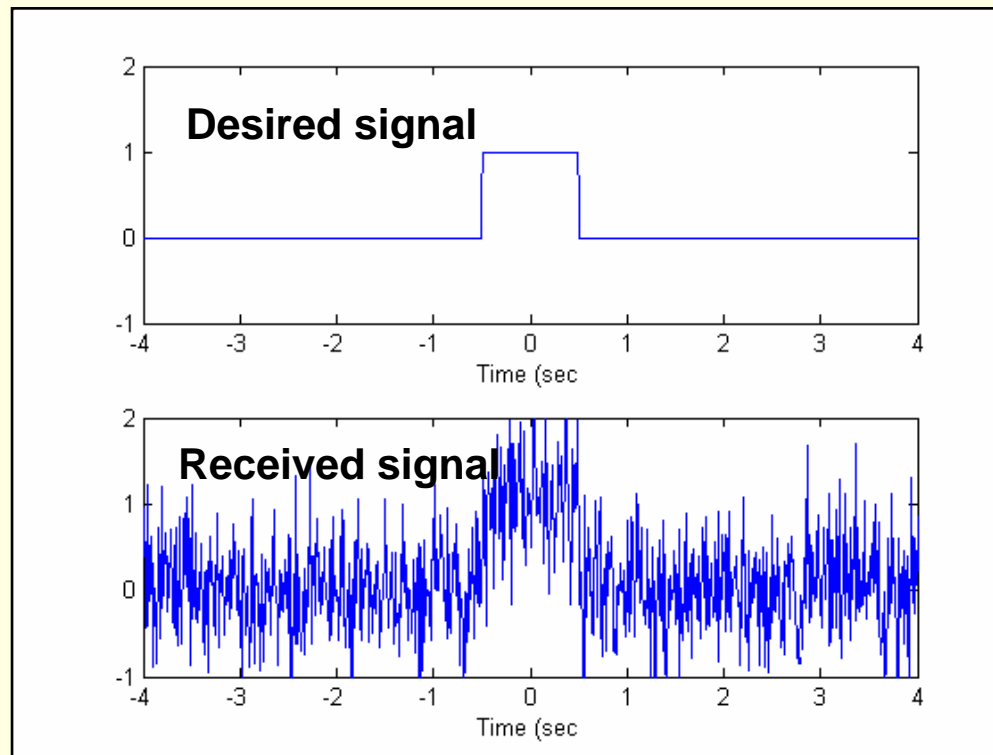


- A major use of filters is the elimination of noise.
- Noise typically has a much larger bandwidth than the signal of interest.
 - Filtering the received signal with a bandpass or lowpass filter can reduce the amount of noise
- We typically consider *additive noise* where the received signal $r(t)$ is equal to the desired signal $x(t)$ plus noise $n(t)$
 - $r(t) = x(t) + n(t)$

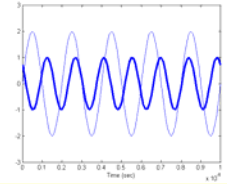
Example 3 (cont.)



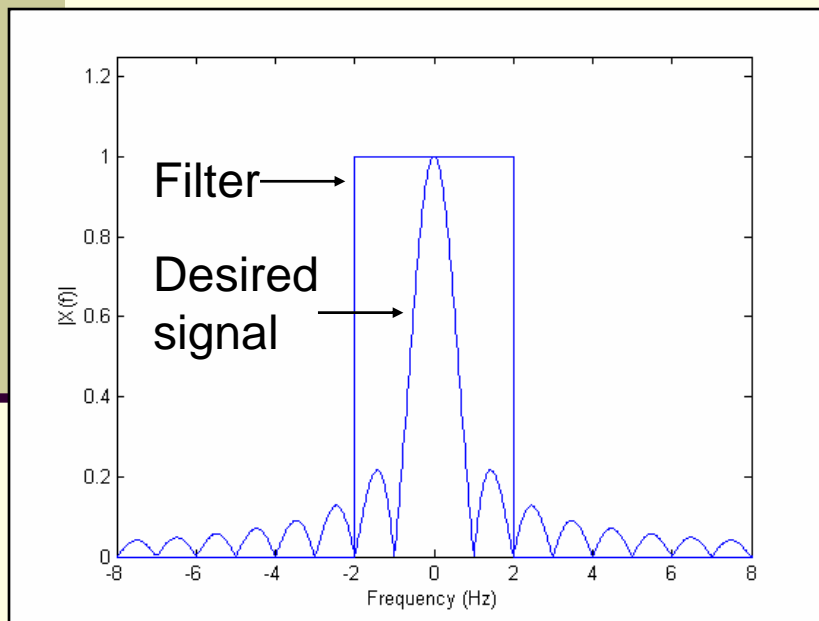
- Consider a square pulse with duration one second that is received with the addition of noise
- The ratio of the received desired signal power to the noise power, signal-to-noise ratio or SNR, is 2



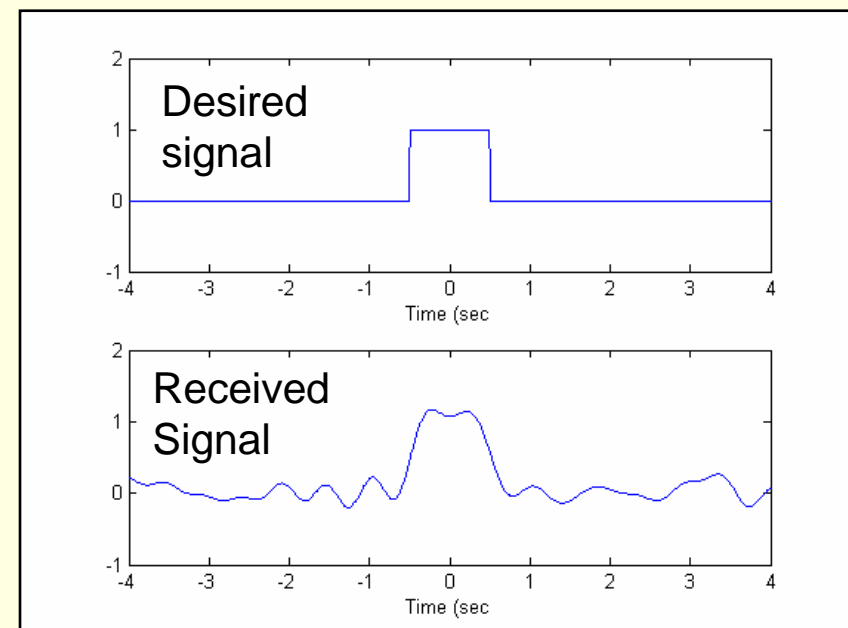
Example 3 (cont.)



- If we filter the signal with an ideal LPF with bandwidth $B = 2\text{Hz}$, we know that we will introduce some distortion to the desired signal, but we can also eliminate much of the noise

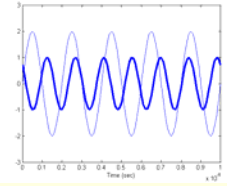


Frequency Domain

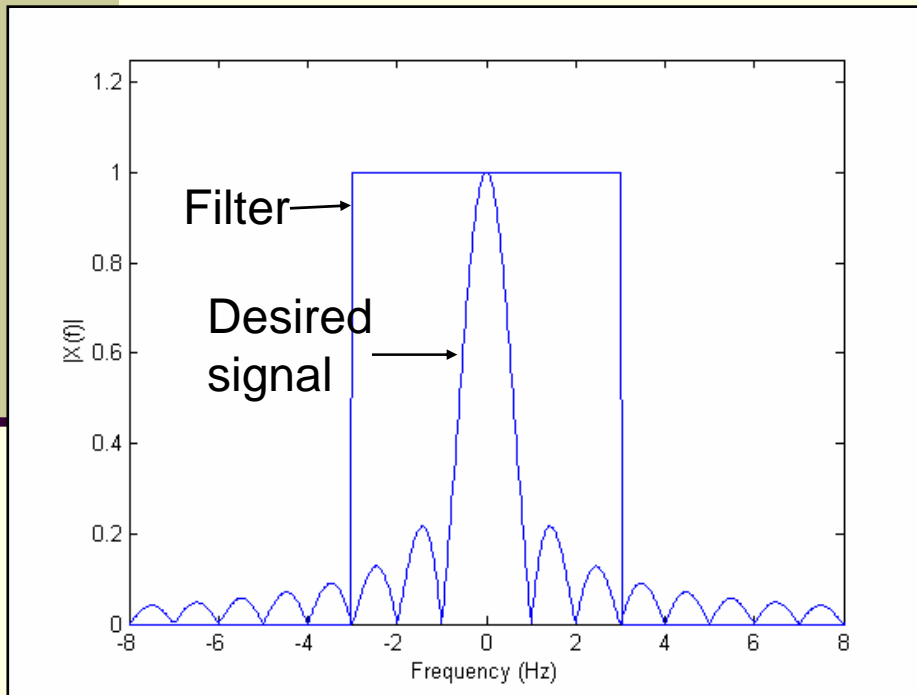


Time Domain

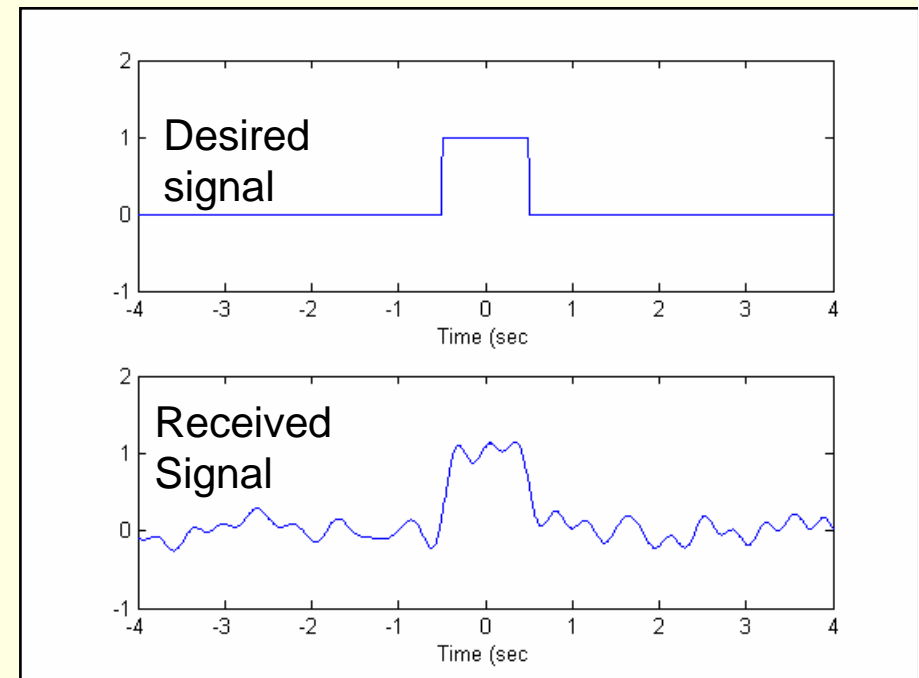
Example 3 (cont.)



- We know that increasing the bandwidth to $B = 3\text{Hz}$ will reduce the amount of distortion to the original signal
- However, it also lets more noise in

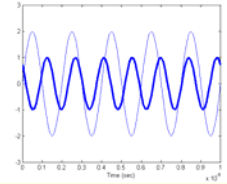


Frequency Domain

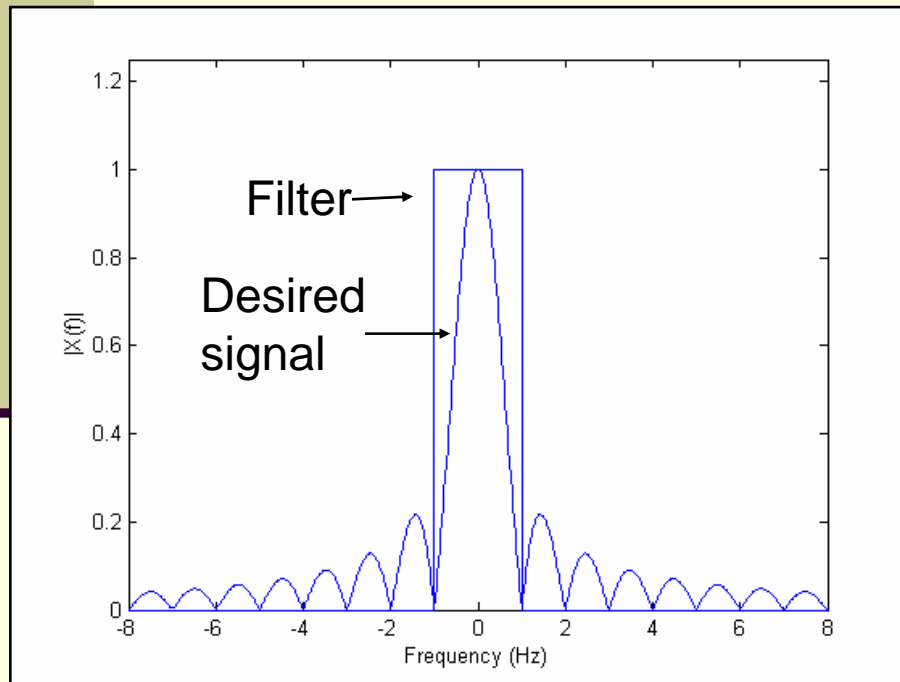


Time Domain

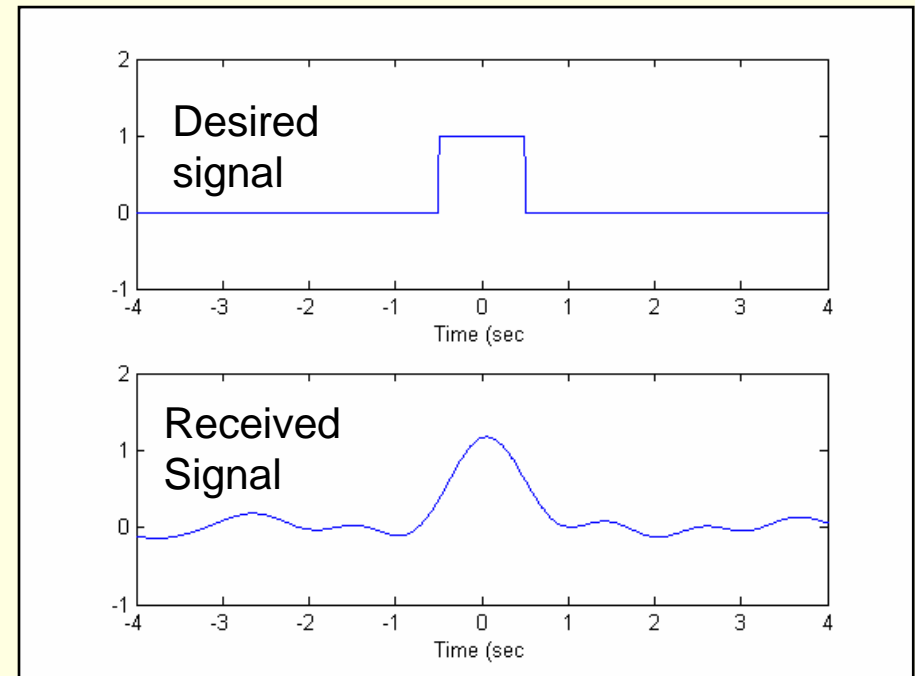
Example 3 (cont.)



- If we decrease the bandwidth to $B = 1\text{ Hz}$ will reduce the amount of noise in the received signal
- However, it introduces more distortion to the desired signal

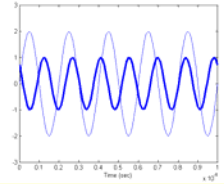


Frequency Domain



Time Domain

Summary



- In this lecture we have examined the application of the Fourier Transform to system analysis, specifically through the system frequency response and through the application of filters
- We defined several ideal filters:
 - LPF, HPF, BPF
- Ideal filters require a non-causal impulse response
 - Real filters must use a delay or non-ideal response
 - Next class we will discuss practical filters