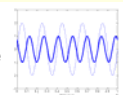



ECE 2704
 Signals and Systems
 Spring 2006

Instructor: Dr. R. Michael Buehrer
 Lecture #8: An Introduction to the
 Fourier Series

Overview

- Today we introduce the concept of the Continuous Time Fourier Series (CTFS)
- What to read – Section 4.1-4.2 in the text
- We will discuss the two versions of the CTFS
 - In terms of complex sinusoids
 - Applies to both real and complex signals
 - In terms of real sinusoids
 - Applies to real signals

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Motivation

- If a system is linear and time invariant (LTI), then when the input is a weighted complex sinusoid, the output is also a complex sinusoid at the same frequency but (in general) with a different weighting.

$$Ae^{j\omega x} \xrightarrow{\mathcal{H}} Be^{j\omega x}$$

- If we can represent a signal as a weighted sum of complex sinusoids, then for an LTI system we can represent the output as the sum of weighted complex sinusoids.
 - Thus, we wish to show that signals can be represented as a sum of complex sinusoids

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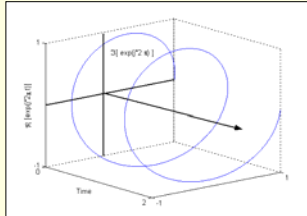
Complex Sinusoids



- Euler's identity defines a complex sinusoid as $e^{jt} = \cos t + j \sin t$

- Real sines and cosines can be written as

$$\cos t = \frac{e^{jt} + e^{-jt}}{2} \quad \sin t = \frac{e^{jt} - e^{-jt}}{2j}$$



Plot of $\exp(j2\pi t)$

Example



- Consider the waveform

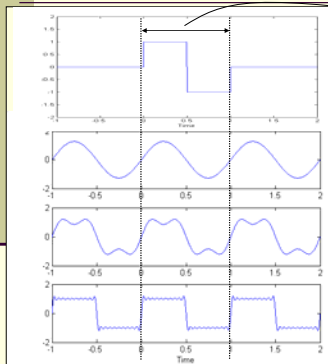
$$x(t) = \text{rect}\left(2t - \frac{1}{2}\right) - \text{rect}\left(2t - \frac{3}{2}\right)$$

- Let us attempt to model this over the time frame $0 < t < 1$ with sinusoids of period 1 and integer multiples of 1:

$$\hat{x}(t) = \frac{4}{\pi} \left(\sin(2\pi t) + \frac{1}{3} \sin(3 * 2\pi t) + \frac{1}{5} \sin(5 * 2\pi t) + \frac{1}{7} \sin(7 * 2\pi t) \dots \right)$$

- Note that we only care about the approximation over the time interval of interest

Example (cont.)



Interval of interest

$$x(t) = \text{rect}\left(2t - \frac{1}{2}\right) - \text{rect}\left(2t - \frac{3}{2}\right)$$

$$\hat{x}(t) = \frac{4}{\pi} \sin(2\pi t)$$

$$\hat{x}(t) = \frac{4}{\pi} \left(\sin(2\pi t) + \frac{1}{3} \sin(3 * 2\pi t) \right)$$

$$\hat{x}(t) = \frac{4}{\pi} \left(\sum_{i=1}^{\infty} \frac{1}{2i-1} \sin(2\pi(2i-1)t) \right)$$

Example (cont.)



- This example shows us that for a specific time interval it appears possible to represent a signal using a sum of sinusoids.
- However, how do we determine the frequency of those sinusoids and how do we determine the weights of the sinusoids?
 - A reasonable guess for frequencies would be multiples of a *fundamental* frequency which is equal to $1/T_f$ where T_f is the length of the interval being examined
 - This will be particularly useful when we examine *periodic* signals
 - We still need to determine the weights
- Further we may want to use both sines and cosines

Signal Representation



- Let's first assume that a signal $x(t)$ can be represented over a time interval $t_0 < t < t_0 + T_f$ as a linear combination of complex sinusoids in the form

$$x_f(t) = \sum_{k=-\infty}^{\infty} X[k] e^{j2\pi k f_f t}$$

where $f_f = 1/T_f$ is the fundamental frequency and

$$x(t) = x_f(t) \text{ for } t_0 < t < t_0 + T_f$$

- We will determine later when this can be safely assumed.
- Note that we are representing the signal $x(t)$ over a finite time interval, not over all time.
- This is defined as the *Continuous Time Fourier Series* or CTFS

Finding the coefficients



- The CTFS representation requires us to determine the coefficients $X[k]$
- Now, since $x(t) = x_f(t)$ $t_0 < t < t_0 + T_f$, then we can write

$$x(t) = \sum_{k=-\infty}^{\infty} X[k] e^{j2\pi k f_f t} \quad t_0 < t < t_0 + T_f$$

- Now, let's multiply both sides by $\exp(-j2\pi q f_f t)$ where q is an integer

$$\begin{aligned} x(t) e^{-j2\pi q f_f t} &= \sum_{k=-\infty}^{\infty} X[k] e^{j2\pi k f_f t} e^{-j2\pi q f_f t} \\ &= \sum_{k=-\infty}^{\infty} X[k] e^{j2\pi(k-q) f_f t} \quad t_0 < t < t_0 + T_f \end{aligned}$$

Finding the coefficients (cont.)



- Now, integrating both sides over $t_0 < t < t_0 + T_f$

$$\int_{t_0}^{t_0+T_f} x(t) e^{-j2\pi q f_s t} dt = \int_{t_0}^{t_0+T_f} \left[\sum_{k=-\infty}^{\infty} X[k] e^{j2\pi(k-q)f_s t} \right] dt$$

- Since k and t are independent we can interchange summation and integration

$$\int_{t_0}^{t_0+T_f} x(t) e^{-j2\pi q f_s t} dt = \sum_{k=-\infty}^{\infty} X[k] \int_{t_0}^{t_0+T_f} e^{j2\pi(k-q)f_s t} dt \quad \text{Equation A}$$

- Using Euler's identity

$$\int_{t_0}^{t_0+T_f} x(t) e^{-j2\pi q f_s t} dt = \sum_{k=-\infty}^{\infty} X[k] \int_{t_0}^{t_0+T_f} [\cos(2\pi(k-q)f_s t) + j \sin(2\pi(k-q)f_s t)] dt$$

Continuing...



$$\int_{t_0}^{t_0+T_f} x(t) e^{-j2\pi q f_s t} dt = \sum_{k=-\infty}^{\infty} X[k] \int_{t_0}^{t_0+T_f} [\cos(2\pi(k-q)f_s t) + j \sin(2\pi(k-q)f_s t)] dt$$

- $k-q$ is an integer. For $k \neq q$ we are integrating $\cos(2\pi[k-q]f_s t)$ and $\sin(2\pi[k-q]f_s t)$ over exactly $k-q$ periods which results in zero.

- For $k \neq q$

$$\int_{t_0}^{t_0+T_f} [\cos(2\pi(k-q)f_s t) + j \sin(2\pi(k-q)f_s t)] dt = 0$$

- For $k = q$

$$\int_{t_0}^{t_0+T_f} [\cos(0) + j \sin(0)] dt = T_f$$

Continuing...



- Therefore,

$$\sum_{k=-\infty}^{\infty} X[k] \int_{t_0}^{t_0+T_f} e^{-j2\pi(k-q)f_s t} dt = X[q] T_f$$

- Making this substitution into Equation A

$$\int_{t_0}^{t_0+T_f} x(t) e^{-j2\pi q f_s t} dt = X[q] T_f$$

- Solving for $X[q]$:

$$X[q] = \frac{1}{T_f} \int_{t_0}^{t_0+T_f} x(t) e^{-j2\pi q f_s t} dt$$

- Since this is true for $X[q]$, then $X[k]$ is

$$X[k] = \frac{1}{T_f} \int_{t_0}^{t_0+T_f} x(t) e^{-j2\pi k f_s t} dt$$

Continuous Time Fourier Transform



- Thus, if the integral converges, we can represent the signal $x(t)$ exactly, on the time interval $t_0 < t < t_0 + T_f$ by

$$x_F(t) = \sum_{k=-\infty}^{\infty} X[k] e^{j2\pi k f_s t}$$

where

$$X[k] = \frac{1}{T_f} \int_{t_0}^{t_0 + T_f} x(t) e^{-j2\pi k f_s t} dt$$

- Note that if the above integral does not converge, the CTFS cannot be found over the region of interest.
- We can represent the relationship between $x(t)$ and $X[k]$ as

$$x(t) \overset{FS}{\leftrightarrow} X[k]$$

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Dirichlet Conditions



- If the integral

$$X[k] = \frac{1}{T_f} \int_{t_0}^{t_0 + T_f} x(t) e^{-j2\pi k f_s t} dt$$

diverges, the CTFS cannot be found for the signal over the region of interest.

- This condition is necessary but not sufficient.
- Sufficient conditions on the existence of the CTFS are known as the *Dirichlet Conditions*:

- The signal must be absolutely integrable over the time $t_0 < t < t_0 + T_f$

$$\int_{t_0}^{t_0 + T_f} |x(t)| dt < \infty$$
- The signal must have a finite number of maxima and minima in the time $t_0 < t < t_0 + T_f$
- The signal must have a finite number of discontinuities in the time $t_0 < t < t_0 + T_f$

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Complex Conjugate of $x(t)$



- The CTFS derived previously holds for any signal which satisfies the Dirichlet conditions
 - $x_F(t) = \sum_{k=-\infty}^{\infty} X[k] e^{j2\pi k f_s t}$ applies to both real signals and complex signals
- Consider the complex conjugate of a signal $x_F(t)$
- By conjugating both sides of our expression:

$$x_F^*(t) = \sum_{k=-\infty}^{\infty} X^*[k] e^{-j2\pi k f_s t}$$

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Complex Conjugate (cont.)



■ Simplifying:

$$\begin{aligned}
 x_F^*(t) &= \sum_{k=-\infty}^{\infty} X^*[k] e^{-j2\pi k f_s t} \\
 &= \sum_{k=-\infty}^{\infty} X^*[-k] e^{j2\pi k f_s t} \\
 &= \sum_{k=-\infty}^{\infty} X^*[-k] e^{j2\pi k f_s t}
 \end{aligned}$$

■ Thus, we can say that if

$$x(t) \leftrightarrow X[k] \quad \text{FS}$$

then

$$x^*(t) \leftrightarrow X^*[-k] \quad \text{FS}$$

Real Signals



■ For the important case where the signal is all real
which means that

$$x(t) = x^*(t)$$

■ In this case

$$x_F(t) = x_F^*(t)$$

$$x_F^*(t) = \sum_{k=-\infty}^{\infty} X^*[-k] e^{j2\pi k f_s t} = x_F(t) = \sum_{k=-\infty}^{\infty} X[k] e^{j2\pi k f_s t}$$

which leads to

$$\sum_{k=-\infty}^{\infty} X[k] e^{j2\pi k f_s t} = \sum_{k=-\infty}^{\infty} X^*[-k] e^{-j2\pi(-k) f_s t}$$

meaning that

$$= \sum_{k=-\infty}^{\infty} X^*[-k] e^{j2\pi k f_s t}$$

$$X[k] = X^*[-k]$$

Real Signals (cont.)



■ For real signals then we can write the CTFS as

$$x_F(t) = X[0] + \sum_{k=1}^{\infty} [X[k] e^{j2\pi k f_s t} + X^*[k] e^{-j2\pi k f_s t}]$$

which can also be written as

$$\begin{aligned}
 x_F(t) &= X[0] + \sum_{k=1}^{\infty} \{ \text{Re}(X[k]) e^{j2\pi k f_s t} + \text{Re}(X[k]) e^{-j2\pi k f_s t} \\
 &\quad + j \text{Im}(X[k]) e^{j2\pi k f_s t} - j \text{Im}(X[k]) e^{-j2\pi k f_s t} \}
 \end{aligned}$$

Examining the first term:

$$X[0] = \frac{1}{T_F} \int_{t_0}^{t_0+T_F} x(t) dt$$

which is simply the average value of the function (note that it is all real).

Trigonometric CTFS



- Recall that $\cos t = \frac{e^{jt} + e^{-jt}}{2}$ $\sin t = \frac{e^{jt} - e^{-jt}}{2j}$

- Substituting these relationships into the previous equation we have

$$x_F(t) = X[0] + \sum_{k=1}^{\infty} \{ 2 \operatorname{Re}\{X[k]\} \cos(2\pi k f_F t) + 2 \operatorname{Im}\{X[k]\} \sin(2\pi k f_F t) \}$$

which can be re-written as

$$x_F(t) = X_c[0] + \sum_{k=1}^{\infty} \{ X_c[k] \cos(2\pi k f_F t) + X_s[k] \sin(2\pi k f_F t) \}$$

with

$$X_c[k] = 2 \operatorname{Re}\{X[k]\} = 2 \operatorname{Re} \left\{ \frac{1}{T_F} \int_c^{c+T_F} x(t) e^{-j2\pi k f_F t} dt \right\} \quad X_s[k] = -2 \operatorname{Im}\{X[k]\} = -2 \operatorname{Im} \left\{ \frac{1}{T_F} \int_c^{c+T_F} x(t) e^{-j2\pi k f_F t} dt \right\}$$

$$= \frac{2}{T_F} \int_c^{c+T_F} x(t) \cos(2\pi k f_F t) dt \quad = \frac{2}{T_F} \int_c^{c+T_F} x(t) \sin(2\pi k f_F t) dt$$

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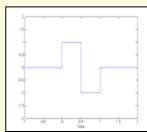
Example



- Let's return to our previous example of

$$x(t) = \operatorname{rect}\left(2t - \frac{1}{2}\right) - \operatorname{rect}\left(2t - \frac{3}{2}\right)$$

- Define the time interval to be $0 < t < 1$



$$x_F(t) = X_c[0] + \sum_{k=1}^{\infty} \{ X_c[k] \cos(2\pi k f_F t) + X_s[k] \sin(2\pi k f_F t) \}$$

$$X_c[0] = \frac{2}{T_F} \int_c^{c+T_F} x(t) \cos(0) dt$$

$$= \frac{2}{T_F} \int_0^{0.5} dt - \frac{2}{T_F} \int_{0.5}^1 dt$$

$$= 0$$

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Example (cont.)



- First note that $f_F = 1/T_F = 1$
- Thus, for the cosine terms we have

$$X_c[k] = 2 \int_0^{0.5} \cos(2\pi k t) dt - 2 \int_{0.5}^1 \cos(2\pi k t) dt$$

$$= \frac{2}{2\pi k} \sin(2\pi k t) \Big|_0^{0.5} - \frac{2}{2\pi k} \sin(2\pi k t) \Big|_{0.5}^1$$

$$= \frac{2}{2\pi k} (\sin(k\pi) - \sin(0)) - \sin(2k\pi) + \sin(k\pi)$$

$$= 0$$

$$X_s[k] = 2 \int_0^{0.5} \sin(2\pi k t) dt - 2 \int_{0.5}^1 \sin(2\pi k t) dt$$

$$= -\frac{2}{2\pi k} (\cos(2\pi k t)) \Big|_0^{0.5} + \frac{2}{2\pi k} (\cos(2\pi k t)) \Big|_{0.5}^1$$

$$= \frac{1}{\pi k} (-\cos(\pi k) + \cos(0)) + \cos(2\pi k) - \cos(\pi k)$$

$$= \frac{1}{\pi k} (2 - 2\cos(\pi k))$$

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Example (cont.)

- Thus, we can write the sine terms as

$$x_s[k] = \frac{1}{\pi k} (2 - 2\cos(\pi k)) = \begin{cases} 0 & k = \text{even} \\ \frac{4}{\pi k} & k = \text{odd} \end{cases}$$

- Thus we have

$$x_F(t) = \sum_{k=1}^{\infty} \frac{2 - 2\cos(\pi k)}{\pi k} \sin(2\pi k f_F t)$$

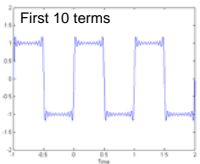
- Since the even terms are zero we can rewrite this as

$$x_F(t) = \frac{4}{\pi} \sum_{k=1}^{\infty} \frac{1}{(2k-1)} \sin(2\pi(2k-1)f_F t)$$

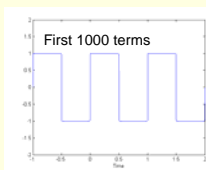
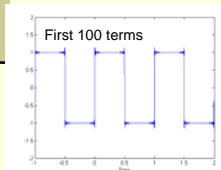
- This is the representation that we examined earlier!

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Example - Plots



- As we continue to add the sinusoidal terms our representation is closer and closer to the original signal *over the interval of interest*.
- When the number of terms goes to infinity the representation is *exact*

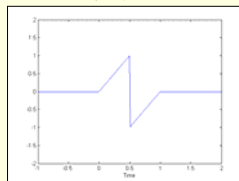


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Example B

- Let's consider a sawtooth waveform

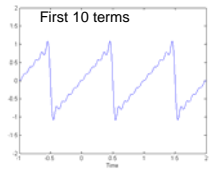
$$x(t) = 2 * \text{ramp}(t) \{u(t) - u(t - \frac{1}{2})\} + (2 * \text{ramp}(t - \frac{1}{2}) - 2) \{u(t - \frac{1}{2}) - u(t - 1)\}$$



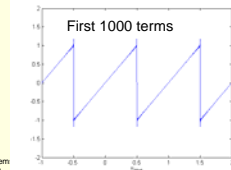
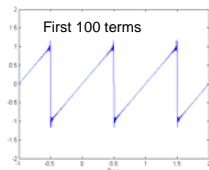
- Define the time interval to be $0 < t < 1$
- $T_F = 1, f_F = 1$

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Example B (final)



- As we continue to add the sinusoidal terms our representation is closer and closer to the original signal *over the interval of interest*.
- When the number of terms goes to infinity the representation is *exact*



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