

3614: Introduction to Communication Systems

Final Exam
December 13, 2006

I pledge that I have neither given nor received any assistance on this exam.

(signed)

Name (print)

Student Number

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1. (20 points) Multiple Choice – Choose the answer which best completes the sentence

1.1 [5 points] The Fourier Transform of a periodic time-domain signal is

- (a) ~~periodic in the frequency domain~~
- (b) discrete in the frequency domain
- (c) ~~continuous in the frequency domain~~
- (d) None of the above

1.2 [5 points] The transfer function of an ideal low pass filter with bandwidth 20kHz can be written as

- (a) $H(f) = \text{rect}\left(\frac{f - f_o}{40000}\right) e^{-j2\pi f T_o}$
- (b) $H(f) = \text{rect}\left(\frac{f}{40000}\right) e^{-j2\pi f T_o}$
- (c) $H(f) = \text{rect}\left(\frac{f}{20000}\right) e^{-j2\pi f T_o}$
- (d) None of the above

1.3 [5 points] If a message bandwidth has a B Hz, the bandwidth of the corresponding vestigial sideband AM signal B_T relates to B as

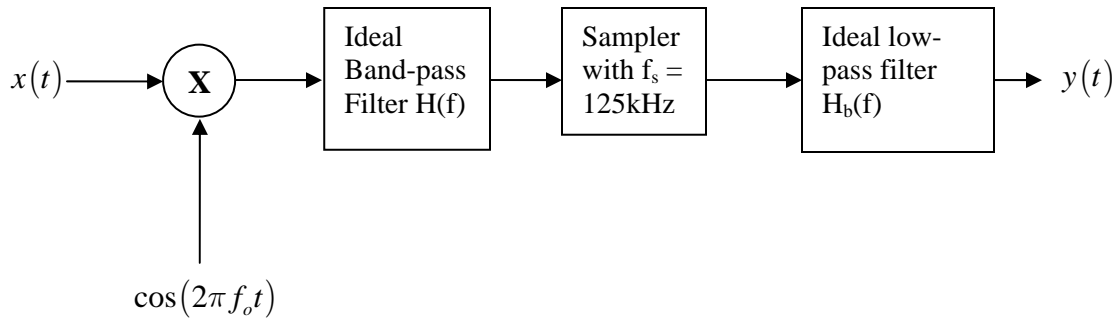
- (a) $B_T < B$ Hz
- (b) $B_T = 2B$ Hz
- (c) $B_T > 2B$ Hz
- (d) $B < B_T < 2B$
- (e) None of the above

1.4 [5 points] The output SNR of an FM system is

- (a) always better than the SNR of a DSBSC system with the same transmit power and message signal
- (b) never better than the SNR of DSBSC system with the same transmit power and message signal
- (c) never better than the SNR of a large carrier AM system with the same transmit power and message signal
- (d) None of the above

2. (25 points) Sampling and Frequency Translation

Consider the following system



where the input signal is

$$x(t) = \cos(2\pi f_c t)$$

with center frequency $f_c = 2\text{MHz}$. The input signal is mixed (multiplied) by a coherent carrier with frequency $f_o = 1.95\text{MHz}$. The resulting signal is passed through an ideal band-pass filter with frequency response

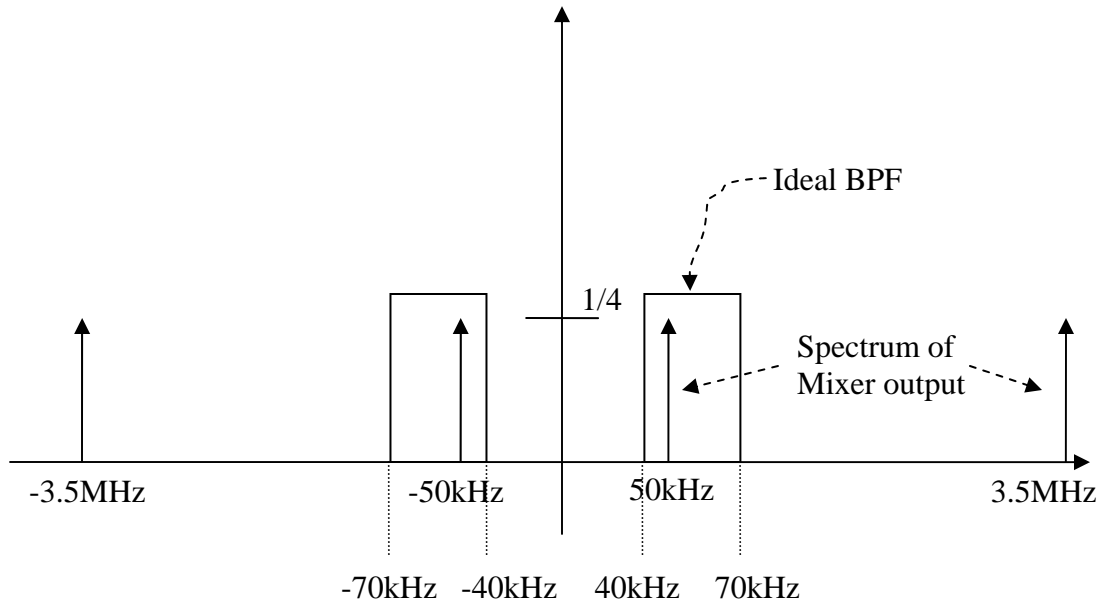
$$H(f) = \text{rect}\left(\frac{f - 40000}{30000}\right) + \text{rect}\left(\frac{f + 40000}{30000}\right)$$

The output of the ideal BPF is input to an ideal sampling device with sampling frequency $f_s = 125\text{kHz}$. The sampled signal is then passed through an ideal low-pass filter with frequency response

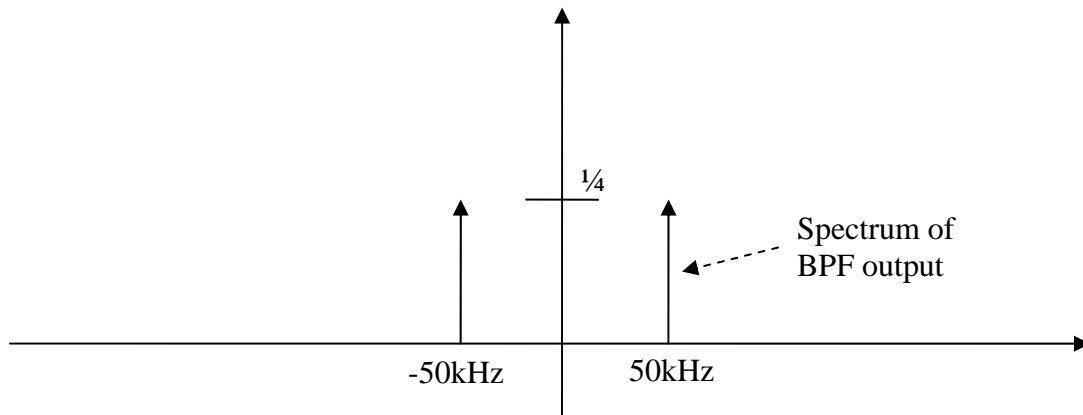
$$H_b(f) = \frac{1}{f_s} \text{rect}\left(\frac{f}{160000}\right)$$

(a) [15 points] Plot the magnitude spectrum $Y(f)$ at the output. Carefully label all curves and axes.

This problem is most easily solved in the frequency domain. The spectrum at the output of the mixer can be drawn as:

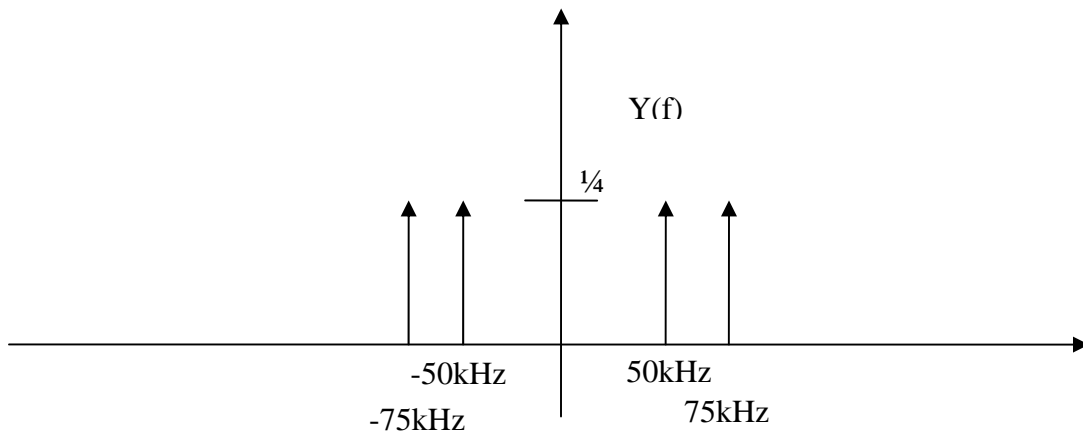
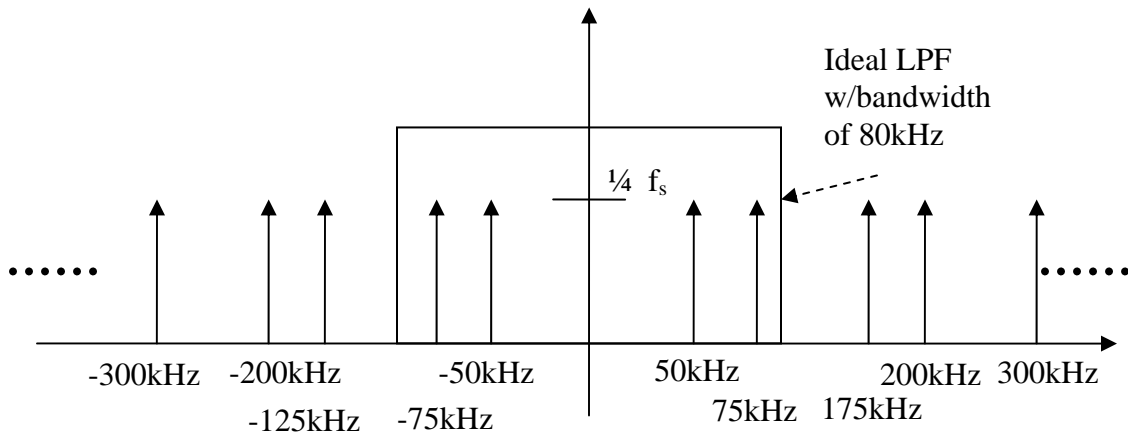
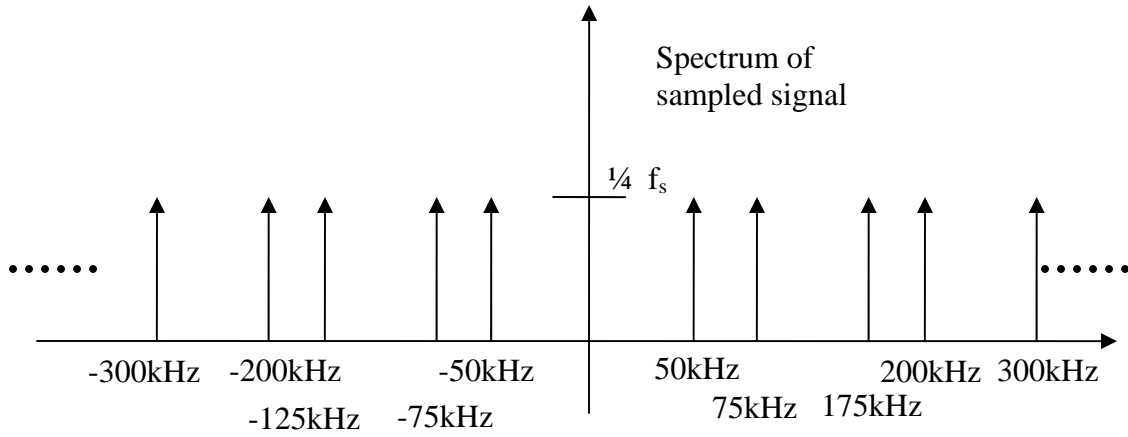


Thus, the spectrum of the signal at the output of the ideal bandpass filter is:



The spectrum of the sampled signal is then drawn as:

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(b) [10 points] Write an equation of the time-domain output $y(t)$.

From the previous plot, we can write the frequency domain result as:

$$Y(f) = \frac{1}{4} \{ \delta(f - 50000) + \delta(f + 50000) + \delta(f - 75000) + \delta(f + 75000) \}$$

Taking the inverse Fourier Transform:

$$Y(f) = \frac{1}{2} \{ \cos(100000\pi t) + \cos(150000\pi t) \}$$

3. (30 points) SNR of AM Systems

If the total received signal power of an amplitude modulated signal is -80dBm, the message bandwidth is 2.5MHz, the message power is 0.1W and the noise power spectral density is $N_o = -170\text{dBm/Hz}$

(a) [10 points] Determine the input (pre-detection) SNR of a DSBSC signal.

The received signal for DSBSC modulation with a message $m(t)$ is

$$s(t) = A_c m(t) \cos(2\pi f_c t)$$

The received signal power can be written as $P = E\{s^2(t)\} = \frac{A_c^2}{2} P_m$ where P_m is the average message power. Thus,

$$10 \log_{10} \left\{ \frac{A_c^2}{2} P_m \right\} = -80 \text{dBm}$$

The pre-detection SNR for DSBSC is

$$SNR_{pre} = \frac{A_c^2}{2} P_m \frac{1}{2N_o W}$$

or in dB:

$$\begin{aligned} SNR_{pre} (dB) &= 10 \log_{10} \left\{ \frac{A_c^2}{2} P_m \right\} - 10 \log_{10} \{2\} - 10 \log_{10} \{W\} - 10 \log_{10} \{N_o\} \\ &= -80 \text{dBm} - 3.01 - 10 \log_{10} \{2.5 * 10^6\} - (-170 \text{dBm} / \text{Hz}) \\ &= 23 \text{dB} \end{aligned}$$

(b) [10 points] Determine the output (post-detection) SNR of a DSBSC signal.

There is a 3dB increase in the SNR at the output of the DSBSC detector. Thus,

$$\begin{aligned} SNR_{pre} (dB) &= SNR_{post} (dB) + 3.01 \\ &= 26 \text{dB} \end{aligned}$$

(c) [10 points] Determine the output (post-detection) SNR if the signal is a large carrier AM signal with amplitude sensitivity $k_a = 0.75$.

The large-carrier AM signal is written as

$$s(t) = A_c (1 + k_a m(t)) \cos(2\pi f_c t)$$

The power is thus $P = E\{s^2(t)\} = \frac{A_c^2}{2}(1 + k_a^2 P_m)$ where P_m is the average message power. Thus,

$$10 \log_{10} \left\{ \frac{A_c^2}{2} (1 + k_a^2 P_m) \right\} = -80 \text{ dBm}$$

$$SNR_{pre} = \frac{A_c^2}{2} (1 + k_a^2 P_m) \frac{1}{2N_o W}$$

$$\begin{aligned} SNR_{pre} (dB) &= 10 \log_{10} \left\{ \frac{A_c^2}{2} (1 + k_a^2 P_m) \right\} - 10 \log_{10} \{2\} - 10 \log_{10} \{W\} - 10 \log_{10} \{N_o\} \\ &= -80 \text{ dBm} - 3.01 - 10 \log_{10} \{2.5 * 10^6\} - (-170 \text{ dBm} / \text{Hz}) \\ &= 23 \text{ dB} \end{aligned}$$

$$\begin{aligned} SNR_{post} (dB) &= SNR_{pre} (dB) + 10 \log_{10} \left(\frac{2k_a^2 P_m}{1 + k_a^2 P_m} \right) \\ &= 23 + 10 \log_{10} \left(\frac{2 * 0.75^2 * 0.1}{1 + 0.75^2 * 0.1} \right) \\ &= 13.3 \text{ dB} \end{aligned}$$

4. (20 points) Consider a message signal

$$m(t) = \text{rect}\left(\frac{t - i/10}{0.0005}\right)$$

where a_i is a random value equally probable on $\{-1, +1\}$.

(a) [10 points] Plot the magnitude spectrum of a large carrier AM signal if the carrier frequency is $f_c = 150\text{kHz}$ and $k_a = 0.25$.

For large carrier AM:

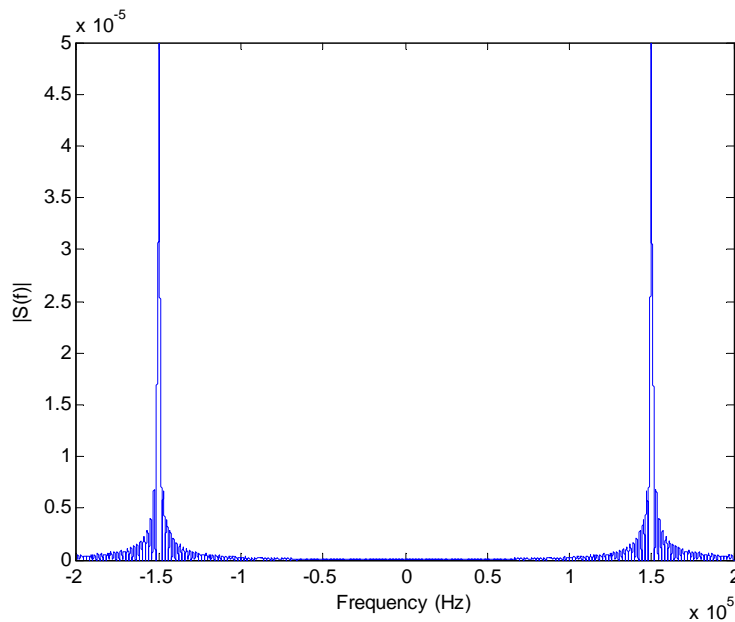
$$S(f) = \frac{A_c}{2} \left[k_a M(f - f_c) + k_a M(f + f_c) + \delta(f - f_c) + \delta(f + f_c) \right]$$

The spectrum of the message signal is

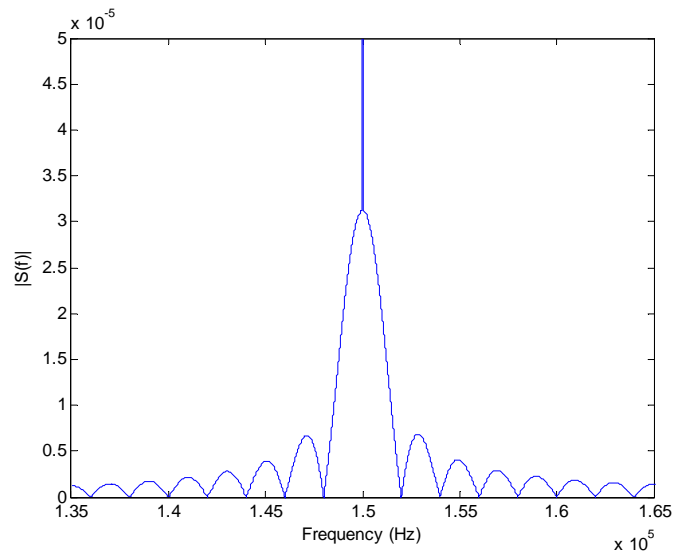
$$M(f) = 0.05 \text{sinc}\left(\frac{f}{20}\right)$$

Thus, we have

$$S(f) = \frac{A_c}{2} \left[0.25 * 0.05 \text{sinc}\left(\frac{f - f_c}{20}\right) + 0.25 * 0.05 \text{sinc}\left(\frac{f + f_c}{20}\right) + \delta(f - f_c) + \delta(f + f_c) \right]$$



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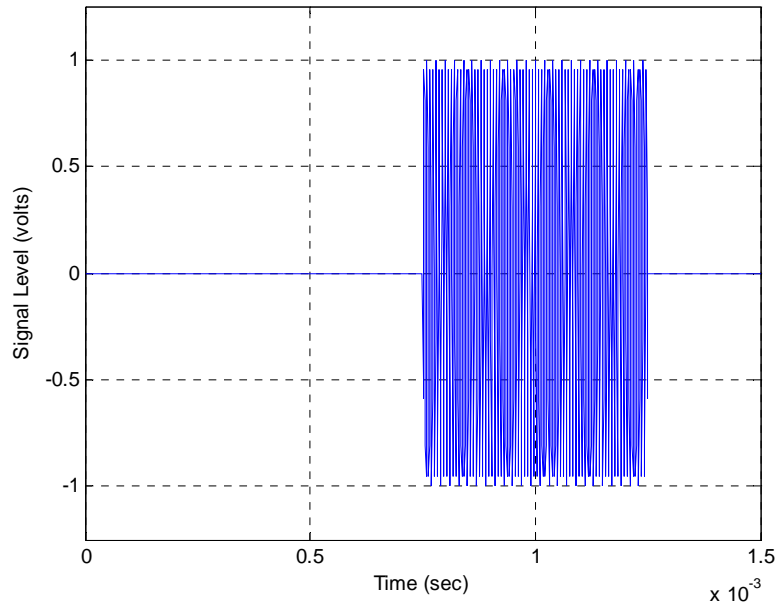


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(b) [10 points] Plot the time-domain function of a DSBSC signal for time $0 < t < 0.0015$. Label all axes and curves.

$$s(t) = A_c m(t) \cos(2\pi f_c t)$$

Assuming $A_c = 1$:



5. [25 points] SNR of FM signals

Consider a received signal with total received power $P_r = -90\text{dBm}$, noise power spectral density $N_o = -165\text{dBm/Hz}$, a message bandwidth of 10kHz and a normalized message power of 0.2W .

(a) [10 points] If the threshold input SNR is 20dB , determine the maximum SNR attainable for an FM system.

The received power for an FM signal is $P = E\{s^2(t)\} = \frac{A_c^2}{2}$, thus,

$10\log_{10}\left\{\frac{A_c^2}{2}\right\} = -80\text{dBm}$. The pre-detection SNR is

$$SNR_{pre} = \frac{A_c^2}{2} \frac{1}{N_o B_T}$$

In dB we have

$$\begin{aligned} SNR_{pre} &= 10\log_{10}\left\{\frac{A_c^2}{2}\right\} - 10\log_{10}\{N_o\} - 10\log_{10}\{B_T\} \\ &= -90\text{dBm} - (-165\text{dBm/Hz}) - 10\log_{10}\{B_T\} \end{aligned}$$

Thus, the pre-detection SNR goes down with input bandwidth. Since, the pre-detection SNR must be greater than 20dB , the input bandwidth is the limiting factor.

The output SNR for FM systems (if above the threshold SNR)

$$SNR_{post} = \frac{3A_c^2 D_f^2 \overline{\left(\frac{m(t)}{V_p}\right)^2}}{2N_o W}$$

As D_f increases the output SNR increases. However, $B_T = 2(D_f + 1)W$. Thus, as D_f increases, the input SNR goes down. To maximize the post-detection SNR, we need to maximize D_f without allowing the pre-detection SNR to go below 20dB . Thus,

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$$SNR_{pre} = 75dBHz - 10\log_{10}\{2(D_f + 1)W\} \geq 20dB$$

$$10\log_{10}\{2(D_f + 1)W\} \leq 55dBHz$$

$$D_f \leq 14.8$$

$$\begin{aligned} SNR_{post} (dB) &= 10\log_{10}(3) + 10\log_{10}\left(\frac{A_c^2}{2}\right) + 10\log_{10}(D_f^2) + 10\log_{10}\left(\left[\frac{m(t)}{V_p}\right]^2\right) - 10\log_{10}(W) - 10\log_{10}(N_o) \\ &= 5dB - 90dBm + 23.4dB - 7dB - 40 - (-165dBm / Hz) \\ &= 56.4dB \end{aligned}$$

(b) [5 points] How does the resulting transmit bandwidth compare to a DSBSC AM system?

$$\begin{aligned} B_T &= 2(D_f + 1)W \\ &= 316kHz \end{aligned}$$

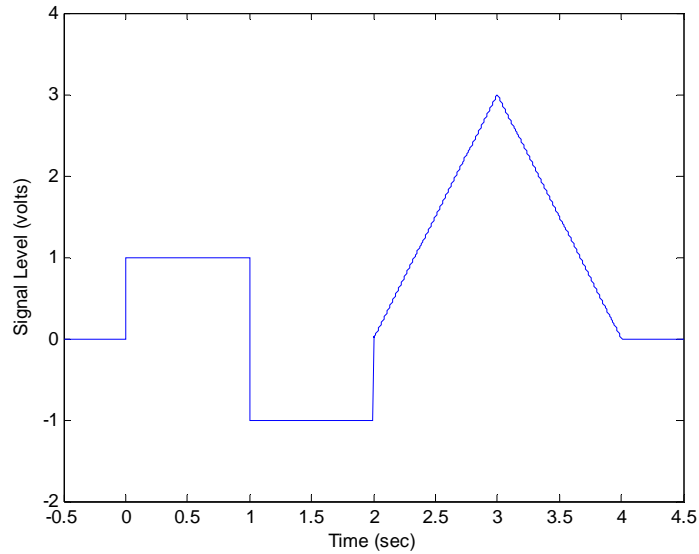
$$\begin{aligned} B_T &= 2W \\ &= 20kHz \end{aligned}$$

(c) [10 points] How does the output SNR compare to an AM system?

$$\begin{aligned} SNR_{post} &= \frac{A_c^2}{2} P_m \frac{1}{N_o B_T} \\ SNR_{post} (dB) &= -90dBm + 165 - 10\log_{10}(2 * 10000) \\ &= -90dBm + 165 - 43 \\ &= 32dB \end{aligned}$$

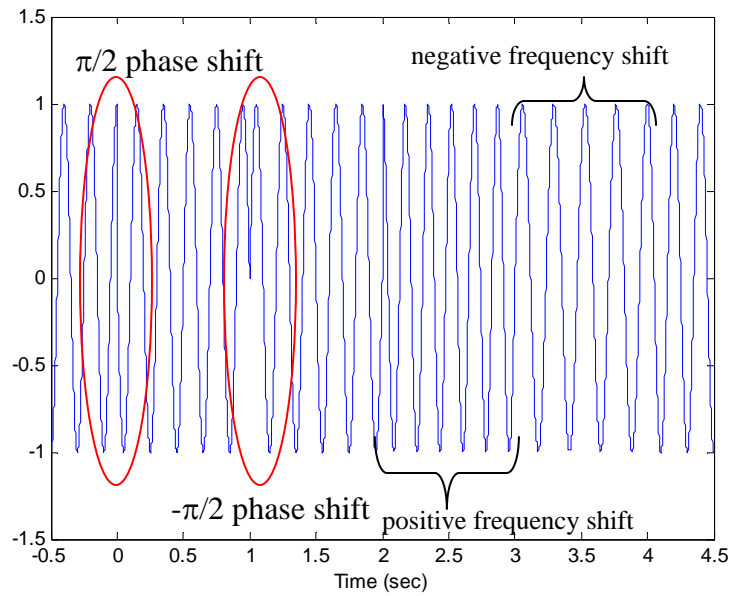
6. (20 points) PM/FM

Consider the message signal



Message Signal for Problem 6

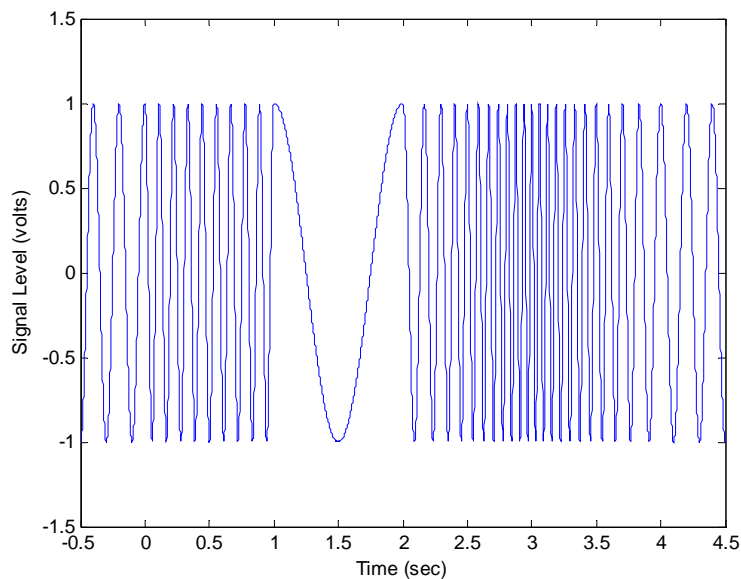
(a) [10 points] Draw the time domain PM signal over the time -0.25 to 4.25 seconds for a carrier frequency of 5 Hz and $k_p = \pi/2$ rad/V.



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Over the interval $-0.5 < t < 0$, the carrier is unmodulated. Over the interval $0 < t < 1$ the signal is phase shifted by positive $\pi/2$ radians. Over the interval $1 < t < 2$ the signal is phase shifted by negative $\pi/2$ radians. Over the interval $2 < t < 3$ the signal phase is changed linearly $3\pi/2$ radians/sec which corresponds to a $3/4$ Hz frequency increase ($f = 5.75\text{Hz}$). Over the interval $3 < t < 4$ the phase of the signal changed linearly $-3\pi/2$ radians/sec which corresponds to a $-3/4$ Hz frequency increase ($f = 4.25\text{Hz}$). For the interval $t > 4$, the carrier is unmodulated.

(b) [10 points] Draw the time domain FM signal over the time -0.25 to 4.25 for a carrier frequency of 5Hz and $k_f = 4\text{Hz/V}$.



Over the interval $-0.5 < t < 0$, the carrier is unmodulated. Over the interval $0 < t < 1$ the signal is frequency shifted by positive 4Hz ($f_c = 9\text{Hz}$). Over the interval $1 < t < 2$ the signal is frequency shifted by negative 4Hz ($f_c = 1\text{Hz}$). Over the interval $2 < t < 3$ the signal frequency is changing linearly 12Hz/sec so that at the end of the interval $f_c = 17\text{Hz}$. Over the interval $3 < t < 4$ the signal frequency is changing linearly -12Hz/sec so that at the end of the interval $f_c = 5\text{Hz}$. For the interval $t > 4$, the carrier is unmodulated.