

# ECE4634 Digital Communications Fall 2007

Instructor: R. Michael Buehrer  
Lecture #32: BERs for  
Non-coherent Demodulation  
and Rayleigh Fading



---

---

---

---

---

---

---

---

## Overview



- In the previous classes we examined the bit error rate performance of PSK, ASK, and FSK using a *coherent* receiver in the presence of AWGN
- Now we would like to look at other important cases including
  - *Non-coherent* receivers in AWGN
  - Coherent receivers in Rayleigh fading

RM Buehrer  
Virginia Tech  
Fall 2007

---

---

---

---

---

---

---

---

## Probability of Error for Noncoherent Reception



- Differentially coherent receivers can detect changes in phase from one symbol to the next
  - Can be used with BPSK, MPSK, and QAM signal constellations
- Noncoherent receivers can only detect the energy of signals in certain frequency bands
  - Used for binary and M-ary FSK receivers

RM Buehrer  
Virginia Tech  
Fall 2007

---

---

---

---

---

---

---

---

## Differential Encoding of Data

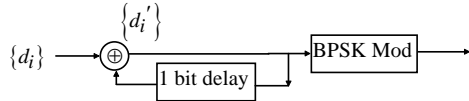


- Consider BPSK modulation;

$$0 \Rightarrow s(t) = \sqrt{2P} \cos(2\pi f_c t) \Big|_0^T$$

$$1 \Rightarrow s(t) = \sqrt{2P} \cos(2\pi f_c t + \pi) \Big|_0^T = -\sqrt{2P} \cos(2\pi f_c t) \Big|_0^T$$

- Differential Encoding Transforms Raw Data:



- Rule: "Change the phase if input data is a 1"

RM Bushner  
Virginia Tech  
Fall 2007

---

---

---

---

---

---

---

---

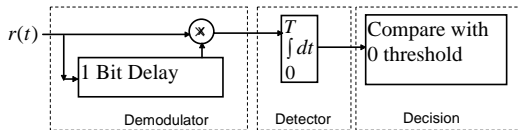
---

---

## Differential Reception



- We can think of differential reception as using the noisy version of the received signal as its phase reference for the correlation operation:



RM Bushner  
Virginia Tech  
Fall 2007

---

---

---

---

---

---

---

---

---

---

## Probability of Error for Differential Receivers



- Standard BPSK:  $P_e = P_b = Q\left\{\sqrt{\frac{2E_b}{N_0}}\right\}$ 
  - symbol error probability and bit error probability the same for binary case
- Standard QPSK

$$P_b = Q\left\{\sqrt{\frac{2E_b}{N_0}}\right\}$$

- DBPSK with Differential Demodulation:

$$P_e = P_b = \frac{1}{2} e^{-E_b/N_0}$$

RM Bushner  
Virginia Tech  
Fall 2007

---

---

---

---

---

---

---

---

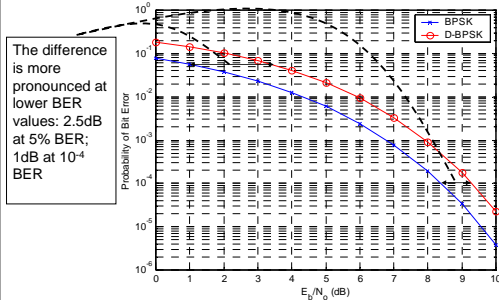
---

---

## Performance Comparison of BPSK and DPSK



DPSK gives about 1.5-2 dB loss from coherent reception.




---

---

---

---

---

---

---

---

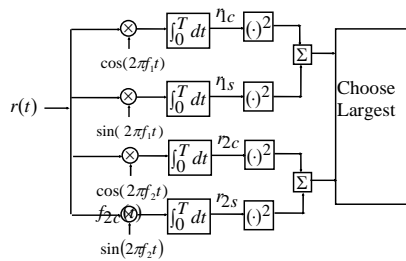
---

---

---

---

## Structure of Optimum Noncoherent Receiver for Binary FSK



RM Buehrer  
Virginia Tech  
Fall 2007

---

---

---

---

---

---

---

---

---

---

---

---

## Probability of Error for Noncoherent FSK



- Probability of error for Binary FSK with coherent reception:

$$P_e = Q\left\{\sqrt{\frac{E_b}{N_0}}\right\}$$

- Probability of error for Binary FSK with noncoherent reception:

$$P_e = \frac{1}{2}e^{-E_b/2N_0}$$

- This results in about a 2 dB loss from coherent reception

RM Buehrer  
Virginia Tech  
Fall 2007

---

---

---

---

---

---

---

---

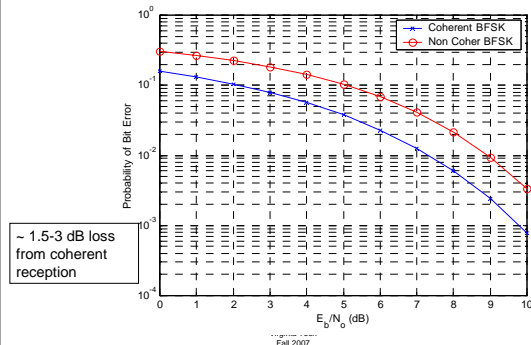
---

---

---

---

## Comparison of Coherent and Noncoherent Binary FSK




---

---

---

---

---

---

---

---

---

---

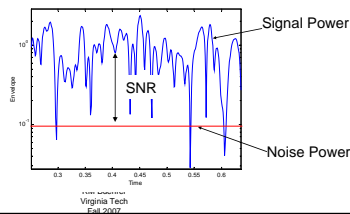
---

---

## Impact of Rayleigh Fading



- In a Rayleigh fading scenario the desired signal power fluctuates due to movement while the noise power remains constant
- As a result the SNR (and thus the performance) fluctuates




---

---

---

---

---

---

---

---

---

---

---

---

## Performance of BPSK in Rayleigh Fading



- Consider the received baseband signal

$$r(t) = \gamma s(t) + n(t)$$

where  $\gamma$  is a constant complex Gaussian value

- If we follow through the same derivation as in standard BPSK (keeping  $\gamma$  as a constant) we arrive at a probability of error given as

$$P_e = Q\left(\sqrt{\frac{2E_b}{N_o} |\gamma|^2}\right)$$

Performance depends on the amplitude of  $\gamma$ . When in a deep fade  $|\gamma|$  can be very small and thus performance very bad.

---

---

---

---

---

---

---

---

---

---

---

---

## Performance

- For a given channel realization the performance is then

$$P_e = Q\left(\sqrt{\frac{2E_b}{N_o}|\gamma|^2}\right) = Q(\sqrt{2\beta})$$

$$\beta = \frac{E_b}{N_o}|\gamma|^2$$

- However, we desire the performance averaged over all channel realizations. Thus, we require the distribution of the random signal-to-noise ratio  $\beta$
- Since the  $\gamma$  is a complex Gaussian random variable,  $\beta$  is a central Chi-Square random variable with two degrees of freedom. The underlying GRV has variance equal to the average signal-to-noise ratio of the channel  $\bar{\beta}$

$$p(\beta) = \frac{1}{\bar{\beta}} e^{-\beta/\bar{\beta}} \quad \beta \geq 0$$

RM Buehner  
Virginia Tech  
Fall 2007




---

---

---

---

---

---

---

---

---

---

## Performance (cont.)

- Substituting  $P_e = \int_0^\infty p(\beta) Q(\sqrt{2\beta}) d\beta$

$$= \int_0^\infty \frac{1}{\bar{\beta}} e^{-\beta/\bar{\beta}} Q(\sqrt{2\beta}) d\beta$$

$$= \frac{1}{2} \left( 1 - \sqrt{\frac{\bar{\beta}}{1+\bar{\beta}}} \right)$$

- In high SNR situations ( $\bar{\beta} \gg 1$ )

$$P_e \approx \frac{1}{2} \left( 1 - \sqrt{\frac{\bar{\beta}}{1+\bar{\beta}}} \right) \approx \frac{1}{2} \left( 1 - \left( 1 - \frac{1}{2\bar{\beta}} \right) \right)$$

$$= \frac{1}{4\bar{\beta}}$$

Recall the binomial series  
 $(1+x)^{-\alpha} = 1 - \alpha x + \frac{\alpha(\alpha+1)x^2}{2!} - \dots$   
 $\frac{\alpha(\alpha-1)(\alpha-2)\dots(\alpha-k+1)x^k}{k!} + \dots$

RM Buehner  
Virginia Tech  
Fall 2007




---

---

---

---

---

---

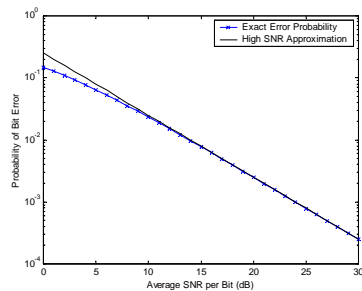
---

---

---

---

## BPSK Performance in Rayleigh Fading



- Rayleigh fading with perfect channel knowledge (we needed to know  $\gamma$ )
- This assumes a *coherent receiver* for BPSK
- Note that the performance is significantly worse than simple AWGN

RM Buehner  
Virginia Tech  
Fall 2007




---

---

---

---

---

---

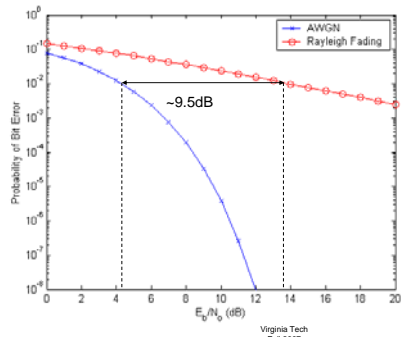
---

---

---

---

## Rayleigh Fading vs. AWGN



- Rayleigh fading results in extremely large performance degradation
- This is particularly true at low BER values

---

---

---

---

---

---

---

---

---

---

## Combating Rayleigh Fading



- The standard method of improving performance in the presence of Rayleigh fading is *diversity*
- Diversity involves having multiple *independent* copies of the received signal
  - If one copy is in a fade, hopefully the other is not
- Diversity can be accomplished using
  - Space – multiple antennas [most common]
  - Time – error correction coding (requires the channel to change in time)
  - Frequency - sending over multiple frequency bands

RM Bushner  
Virginia Tech  
Fall 2007

---

---

---

---

---

---

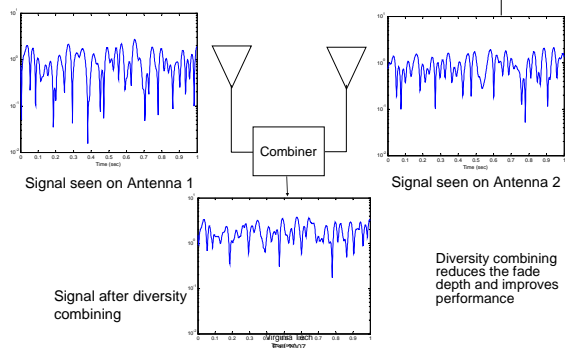
---

---

---

---

## Two Antenna Receive Diversity




---

---

---

---

---

---

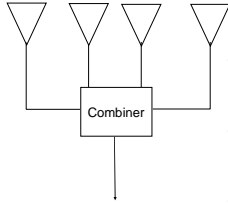
---

---

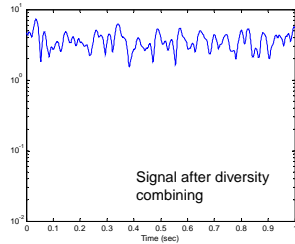
---

---

## Four Antenna Receive Diversity



- Combining more antennas (diversity branches) decreases the fading severity



RM Buehrer  
Virginia Tech  
Fall 2007

---

---

---

---

---

---

---

---

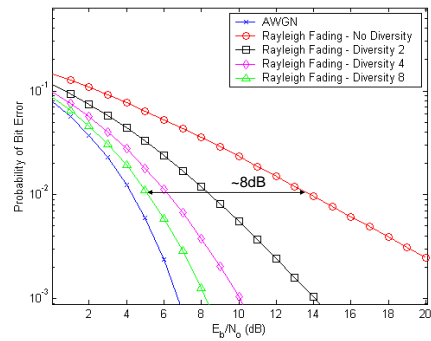
---

---

---

---

## Diversity Improvement



- Large improvements achievable
- Law of diminishing returns applies

---

---

---

---

---

---

---

---

---

---

---

---

## Summary



- We have investigated the performance non-coherent demodulation in an AWGN channel and the performance of coherent demodulation in a Rayleigh fading channel
- Non-coherent reception of BPSK and BFSK results in ~2dB loss over coherent reception
- Similar differences exist for non-coherent  $M$ -ary receivers
- Rayleigh fading results in substantial degradation in BER performance
- Diversity is the standard means of improving performance in Rayleigh fading
- Diversity obeys the law of diminishing returns and can be attained through space (i.e., antennas), frequency or time.

RM Buehrer  
Virginia Tech  
Fall 2007

---

---

---

---

---

---

---

---

---

---

---

---