



# EE 5654 - Digital Communications Spring 2005

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Lecture #16 - Adaptive Algorithms for Equalization  
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# Introduction

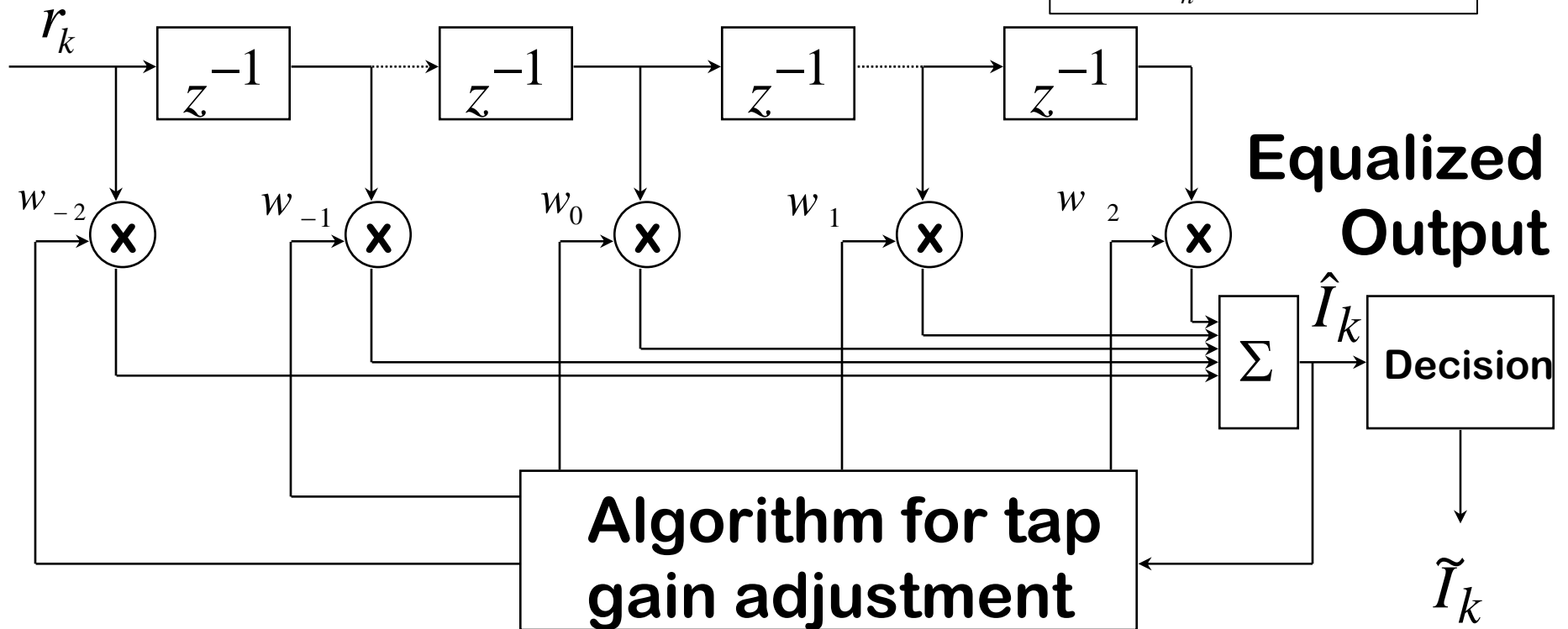
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- Previously we examined equalizers which can improve performance in ISI channels
- In time varying channels, the optimal equalizer coefficients will vary with time and thus must be constantly estimated and updated
- This update technique is called an *adaptive algorithm*
- The algorithm used depends both on the minimization criterion and the update approach

# Linear Transversal Filter

Unequalized Input

$$r_k = \sum_n I_n h_{k-n} + v_k$$





# Linear Transversal Filter Equalizer

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- Complex samples of the received signal input into the equalizer:  $r_k$
- Each element  $z^{-1}$  represents a one sample delay
- There are  $2K+1$  taps on the equalizer where the ISI produces  $K$  samples of overlap in the signal
- The equalizer has  $2K+1$  complex coefficients  $w_k$
- Equalized output samples: 
$$\hat{I}_k = \sum_{j=-K}^K w_j r_{k-j}$$
- Decisions  $\tilde{I}_k$  are made based on  $\hat{I}_k$



# Distortion Criteria

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- Choice of tap coefficients depends on choice of decision criteria.
- Peak distortion criteria leads to “zero-forcing equalizer”
  - Robert Lucky of Bell Labs was famous for his work on this
- Mean Squared Error (MSE) criteria is much more widely used:

$$E\left[\left(I_k - \hat{I}_k\right)^2\right]$$

# Non-adaptive MMSE Solution

- We wish to minimize:  $E \left[ \left| I_k - \sum_{j=-K}^K w_j r_{k-j} \right|^2 \right]$
- In matrix form, it can be shown that the weight vector  $\mathbf{w} = [w_{-K} \ w_{-K+1} \ \dots \ w_{K-1} \ w_K]^T$  which minimizes the MSE is:

$$\mathbf{w} = \mathbf{R}_{rr}^{-1} \mathbf{d}$$

- where  $\mathbf{R}_{rr}$  is the covariance matrix of the input data

$$\mathbf{R}_{rr} = E[\mathbf{r}\mathbf{r}^T]$$

- and  $\mathbf{d}$  is the correlation between the received signal and the information sequence (*i.e.*, the received channel vector)

$$\mathbf{d} = E[\mathbf{r}_k I_k] = \mathbf{h}$$

Note that the vectors are defined over the  $2K+1$  time samples



# More on the Non-adaptive Solution

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- The coefficients  $\mathbf{h}$  represent the channel and are as defined last time (i.e., the ISI and channel of the desired signal after filtering).
- The vector  $\mathbf{w}$  represents the equalizer coefficients
- The  $2K+1 \times 2K+1$  matrix  $\mathbf{R}_{rr}$  represents the covariance matrix of the received signal samples
- The vector  $\mathbf{d}$  represents the correlation between the input signal and the data
- The direct solution requires estimation of  $\mathbf{R}_{rr}$  and  $\mathbf{d}$  followed by inversion of  $\mathbf{R}_{rr}$  and multiplication by  $\mathbf{d}$ . This approach to the MMSE solution is sometimes called Direct Matrix Inversion or DMI



# Problems with Non-adaptive Solution

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- Requires direct computation of a full matrix inversion with every update.
- Explicitly makes use of channel parameters which must be estimated.
- Combined, this results in a very computationally intense approach, especially in a time varying environment

## **Solution: Real-time Adaptive Algorithms**

- Least Mean Square (LMS)
  - Gradient search technique
- Recursive Least Squares (RLS)
  - Estimates matrix inverse

# Least Mean Squares (LMS) Algorithm

- We begin with an initial guess of the solution  $\mathbf{w}_0$ :
- We form a sequence of estimates:  $\mathbf{w}_0, \mathbf{w}_1, \mathbf{w}_2, \dots$  of the solution to the equation:  $\mathbf{R}_{rr}\mathbf{w} = \mathbf{d}$
- The estimates of solution are found from the steepest descent method
- Estimates are updated iteratively:  $\mathbf{w}_{k+1} = \mathbf{w}_k - \Delta \mathbf{G}_k$
- $\Delta$  is a small constant to control the step size of the update
- $\mathbf{G}_k$  is the gradient vector of the error (i.e., the derivative of the MSE wrt the coefficients)

$$\begin{aligned}\mathbf{G}_k &= \mathbf{R}_{rr} \mathbf{w}_k - \mathbf{d} \\ &= -E[e_k \mathbf{r}_k^*]\end{aligned}$$

$$e_k = I_k - \hat{I}_k$$

# Gradient Vector in the LMS Algorithm

- Calculation of the gradient vector requires knowledge of the covariance matrix  $\mathbf{R}_{rr}$  and correlation vector  $\mathbf{d}$ .
- In practice, we use the actual samples as the unbiased estimate of the true gradient.
- This is a stochastic version of the method of steepest descent
- **The LMS Algorithm:**

$$\mathbf{w}_{k+1} = \mathbf{w}_k + \Delta e_k \mathbf{r}_k^*$$

↑  
Unbiased estimate of the gradient



# Interpretation of LMS Algorithm

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- Coefficients are adjusted in small steps to reduce the error in the response of the equalizer to the current set of input samples.
- Coefficients can be updated after each new sample.
- Note that this algorithm implicitly makes use of channel parameters, but only makes use of the received signal samples which we can directly measure.
- Error calculation requires knowledge of the information symbols  $I_k$ , which are in general unknown. Thus, we must send a training sequence periodically. (Can use *decision directed* mode in between training.)



# The Role of Step Size in LMS Algorithm

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- Larger  $\Delta$  means faster convergence to optimum
- Need  $\Delta < 2/\lambda_{\max}$  to guarantee convergence of algorithm, where  $\lambda_{\max}$  is the largest eigenvalue of the covariance matrix  $\mathbf{R}_{rr}$ .
- Larger  $\Delta$  also means that the algorithm will be less able to finely converge to the optimal solution at the end - sometimes called "Excess MSE"
- Too small a choice for  $\Delta$  may prevent the LMS algorithm from adapting to a rapidly-varying channel



# A Key Weakness of the LMS Algorithm

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- Clearly, a good choice of  $\Delta$  must be made to avoid too slow initial convergence or poor final convergence and instability.
- The same step size  $\Delta$  must be used for the update of all coefficients, even though some may be more sensitive to instability and others may be more important to update rapidly
- Solution: Recursive Least Squares Algorithm

# Recursive Least Squares (RLS) Algorithm

- The RLS algorithm implements the optimal solution

$$\mathbf{w} = \mathbf{R}_{rr}^{-1} \mathbf{d}$$

by iteratively estimating  $\mathbf{R}_{rr}$  and  $\mathbf{d}$  and by using the matrix inversion lemma

- Stochastic estimate of  $\mathbf{R}_{rr}$  :

$$\hat{\mathbf{R}}_k = \alpha \hat{\mathbf{R}}_{k-1} + \mathbf{r}_k \mathbf{r}_k^\dagger$$

- Stochastic inverse using matrix inversion lemma:

$$\hat{\mathbf{R}}_k^{-1} = \frac{1}{\alpha} \left\{ \hat{\mathbf{R}}_{k-1}^{-1} - \frac{\hat{\mathbf{R}}_{k-1}^{-1} \mathbf{r}_k \mathbf{r}_k^\dagger \hat{\mathbf{R}}_{k-1}^{-1}}{\alpha + \mathbf{r}_k^\dagger \hat{\mathbf{R}}_{k-1}^{-1} \mathbf{r}_k} \right\}$$



# Recursive Least Squares (RLS) Algorithm

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- We then update the weights according to:

$$\begin{aligned}\mathbf{w}_{k+1} &= \mathbf{R}_{k+1}^{-1} \mathbf{d}_{k+1} \\ &= \frac{1}{\alpha} \left[ \mathbf{R}_k^{-1} - \mathbf{K}_{k+1} \mathbf{r}_k^\dagger \mathbf{R}_k^{-1} \right] \left[ \alpha \mathbf{d}_k + I_k \mathbf{r}_k^* \right] \\ &= \mathbf{w}_k + \mathbf{K}_{k+1} \left[ I_k - \mathbf{r}_k^\dagger \mathbf{w}_k \right] \\ &= \mathbf{w}_k + \mathbf{K}_{k+1} e_k\end{aligned}$$

- Where

$$\mathbf{K}_{k+1} = \frac{\mathbf{R}_k^{-1} \mathbf{r}_k}{\alpha + \mathbf{r}_k \mathbf{R}_k^{-1} \mathbf{r}_k^\dagger}$$



# Recursive Least Squares (RLS) Algorithm (cont.)

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- Coefficients are updated according to:

$$\mathbf{w}_{k+1} = \mathbf{w}_k + \mathbf{K}_{k+1} e_k$$

- In this formulation,  $\mathbf{K}_k$  is called the Kalman gain vector. It is computed according to:

$$\mathbf{K}_{k+1} = \frac{\mathbf{P}_k \mathbf{r}_k}{\alpha + \mathbf{r}_k \mathbf{P}_k \mathbf{r}_k^\dagger}$$

- Where  $\mathbf{P}_k$  is the recursively updated inverse of the correlation matrix

$$\mathbf{P}_{k+1} = \frac{1}{\alpha} [\mathbf{P}_k - \mathbf{K}_{k+1} \mathbf{r}_k \mathbf{P}_k]$$



# Recursive Least Squares (RLS) Algorithm

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- The inverse of the cross-correlation matrix is computed recursively as follows:

$$\mathbf{P}_{k+1} = \frac{1}{\alpha} [\mathbf{P}_k - \mathbf{K}_{k+1} \mathbf{r}_k \mathbf{P}_k]$$

- This recursive method of computing the matrix inversion is the heart of the RLS algorithm.
- Instead of estimating the correlation matrix and inverting it, we estimate the inverse directly
- The coefficient  $0 < \alpha < 1$  is a scalar which we will discuss in a moment.
- An RLS algorithm also exists for the DFE structure



# Interpretations of the RLS Algorithm

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- Rather than minimizing the *average* mean-square error like the LMS, the RLS algorithm is minimizing the actual squared error for the given data sequence.
- The RLS algorithm is an implementation of a more general technique known as Kalman Filtering.
- The Kalman gain replaces the fixed step size in updating the equalizer coefficients.



# The Role of the Weighting Factor

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- The weighting factor  $0 < \alpha < 1$  determines how quickly older data is discarded from the updated estimate.
- Large  $\alpha$  means that older data included in estimate for a longer period of time.
- Small  $\alpha$  means that older data is discounted quickly.



# RLS vs. LMS Algorithms

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- The RLS algorithm converges faster than the LMS algorithm.
- The RLS algorithm is more computationally complex than the LMS algorithm
  - Note that the recursive nature of the RLS algorithm is still simpler than a full matrix inversion operation.
  - Some additional simplifications of the RLS computations are possible.
  - Note that the RLS algorithm operates directly on signal samples.



# Trained vs. Blind Equalization

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- Note that computation of the error  $\varepsilon_k = I_k - \hat{I}_k$  is required for implementation of algorithms.
- Many systems use initial training sequences for which  $I_k$  is known to both transmitter and receiver
  - Example: GSM transmits periodic training sequences as part of each signal frame
- After the algorithm begins to converge, we can switch to “decision-directed” mode:  $\varepsilon_k = \tilde{I}_k - \hat{I}_k$
- Recently, there has been interest in “Blind Equalization”, which requires no training at all



# References on Blind Equalization

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- Y. Sato, "A method of self-recovering equalization for multilevel amplitude modulation systems," *IEEE Trans. Comm.*, pp. 679-682, June 1975.
- D. N. Godard, "Self-recovering equalization and data tracking for two dimensional data communication systems," *IEEE Trans. Comm.*, pp. 1867-1875, Nov. 1980.
- J. Bellini, "Bussgang techniques for blind equalization," *IEEE GLOBECOM '86*, pp. 46.1.1-46.1.7.



# Summary of Adaptive Algorithms

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- Two adaptive algorithms were presented to adapt equalizer coefficients.
- The LMS algorithm is less complex
- The RLS algorithm converges faster
- User-specified parameters determine the speed of convergence for each of these algorithms.

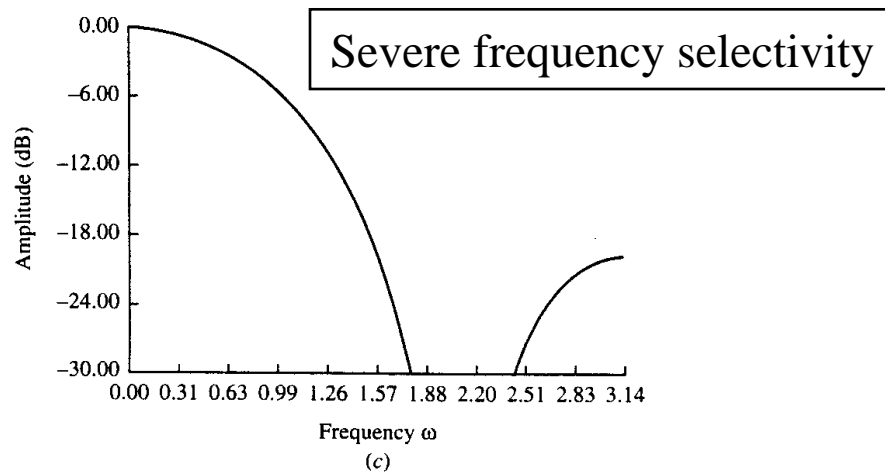
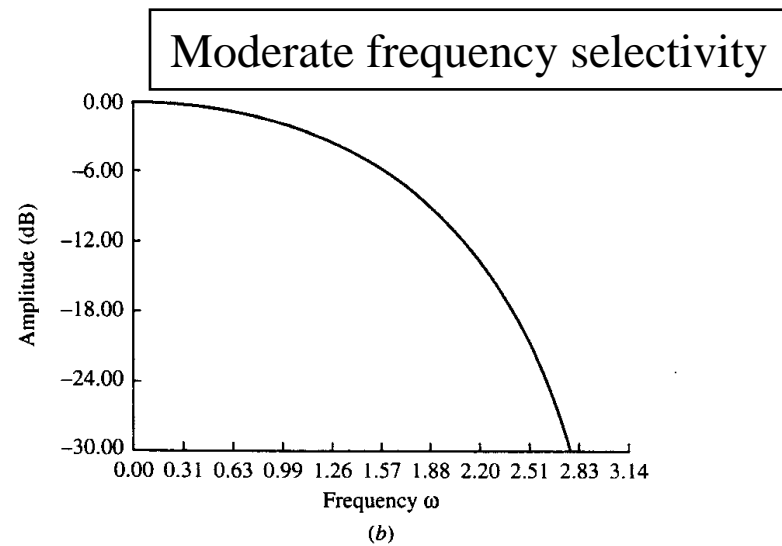
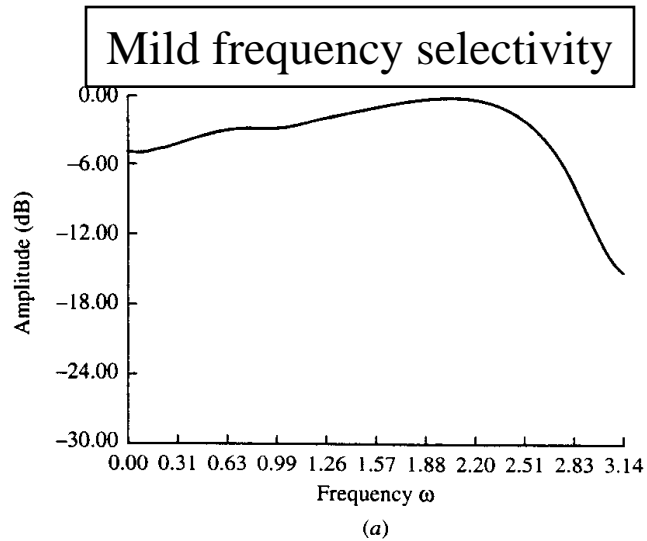


# Performance of Equalizers

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- We are interested in:
  - Steady State Performance
    - Performance After Convergence
    - Largely dependent on structure (provided algorithm converges)
  - Transient Performance
    - Rate of Convergence
    - Dependent on Adaptive Algorithm

# Three Frequency Selective Channels

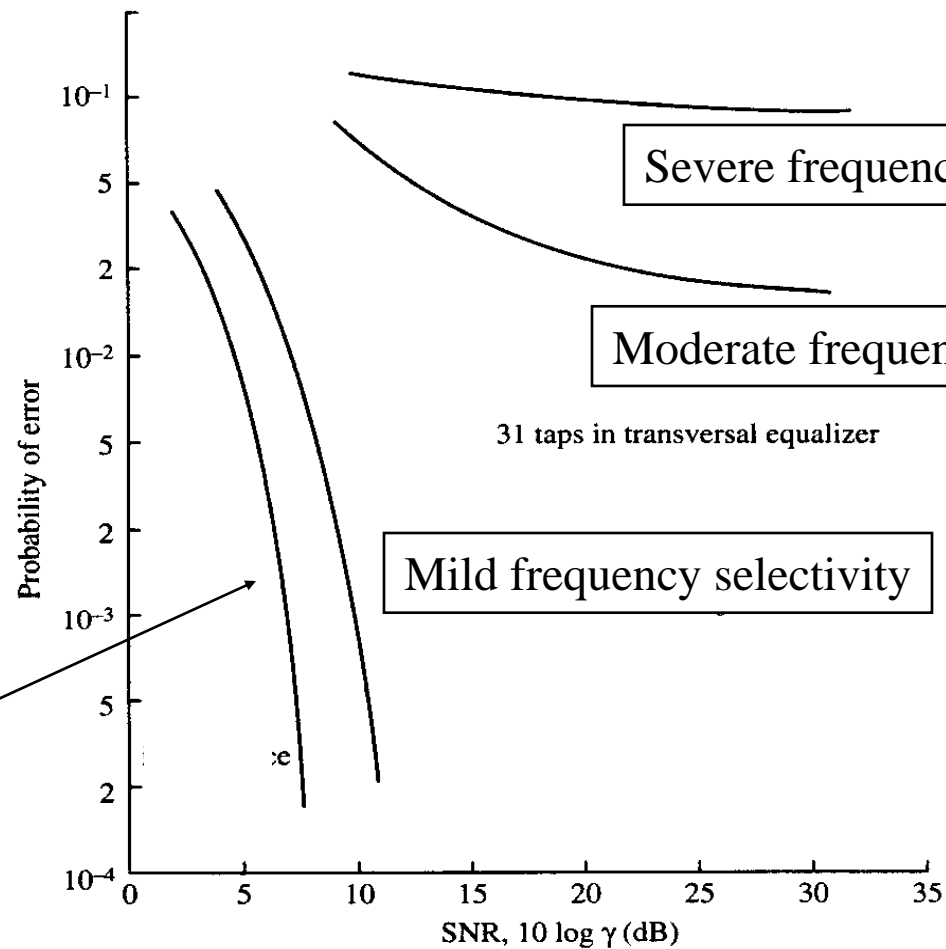


[Proakis 2001]

# Performance of Linear Transversal Filter Equalizer

MMSE  
Equalizer

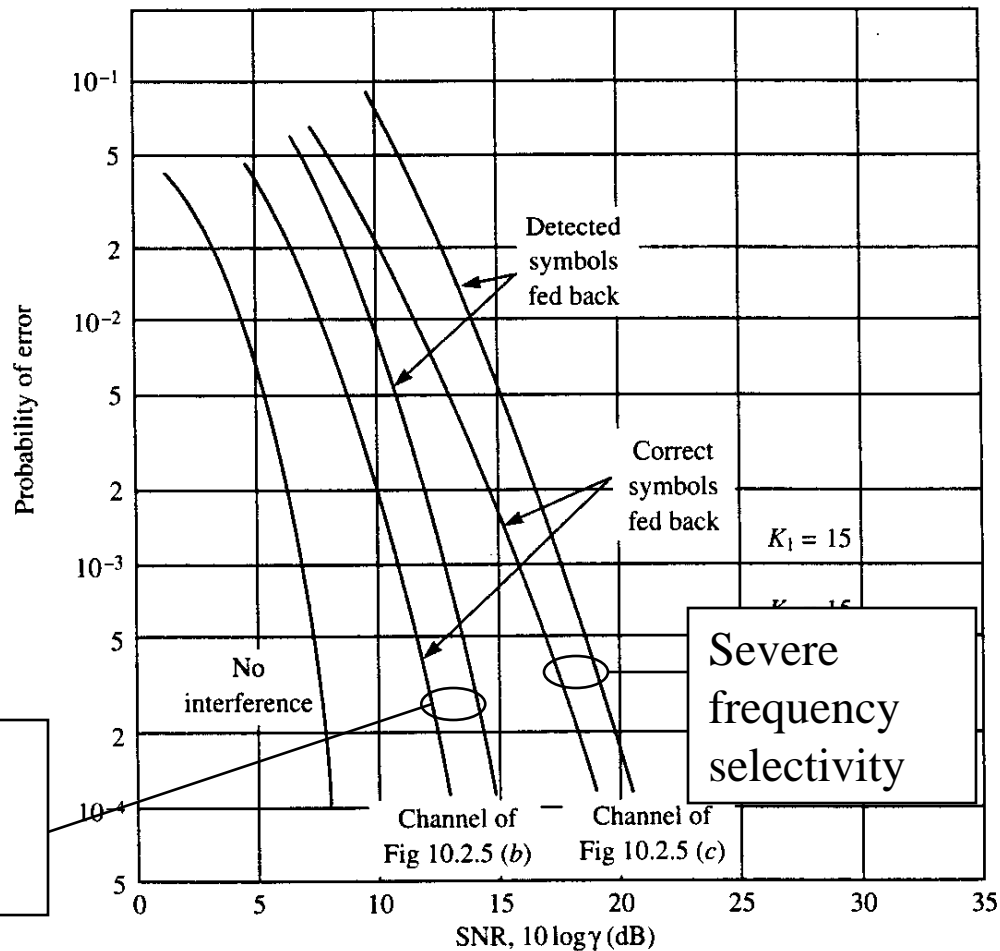
No ISI



[Proakis 2001]

# Performance of DFE with and without Error Propagation

Performance of decision-feedback equalizer with and without error propagation.

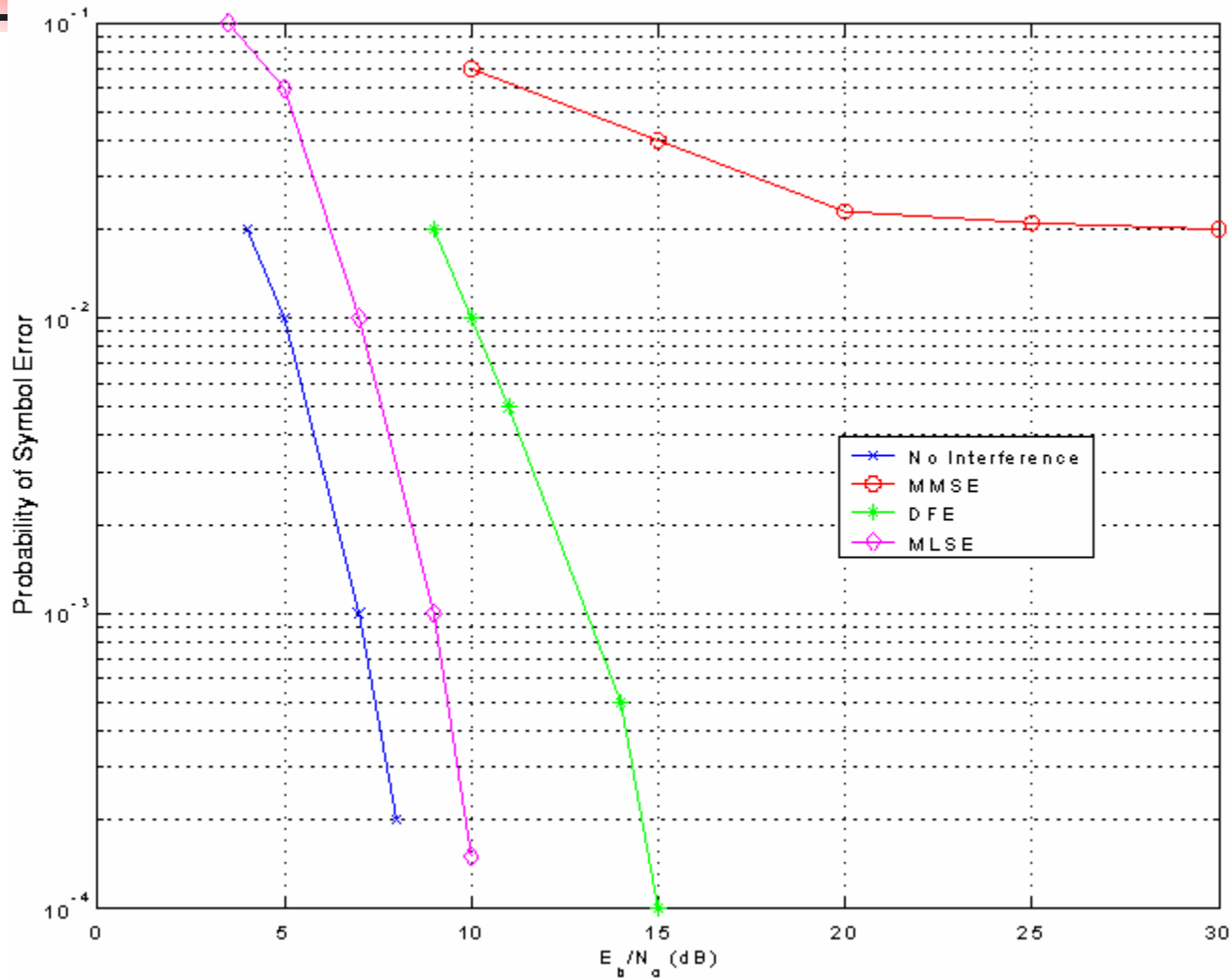


Moderate frequency selectivity

Approximately 2dB penalty due to error propagation

[Proakis 2001]

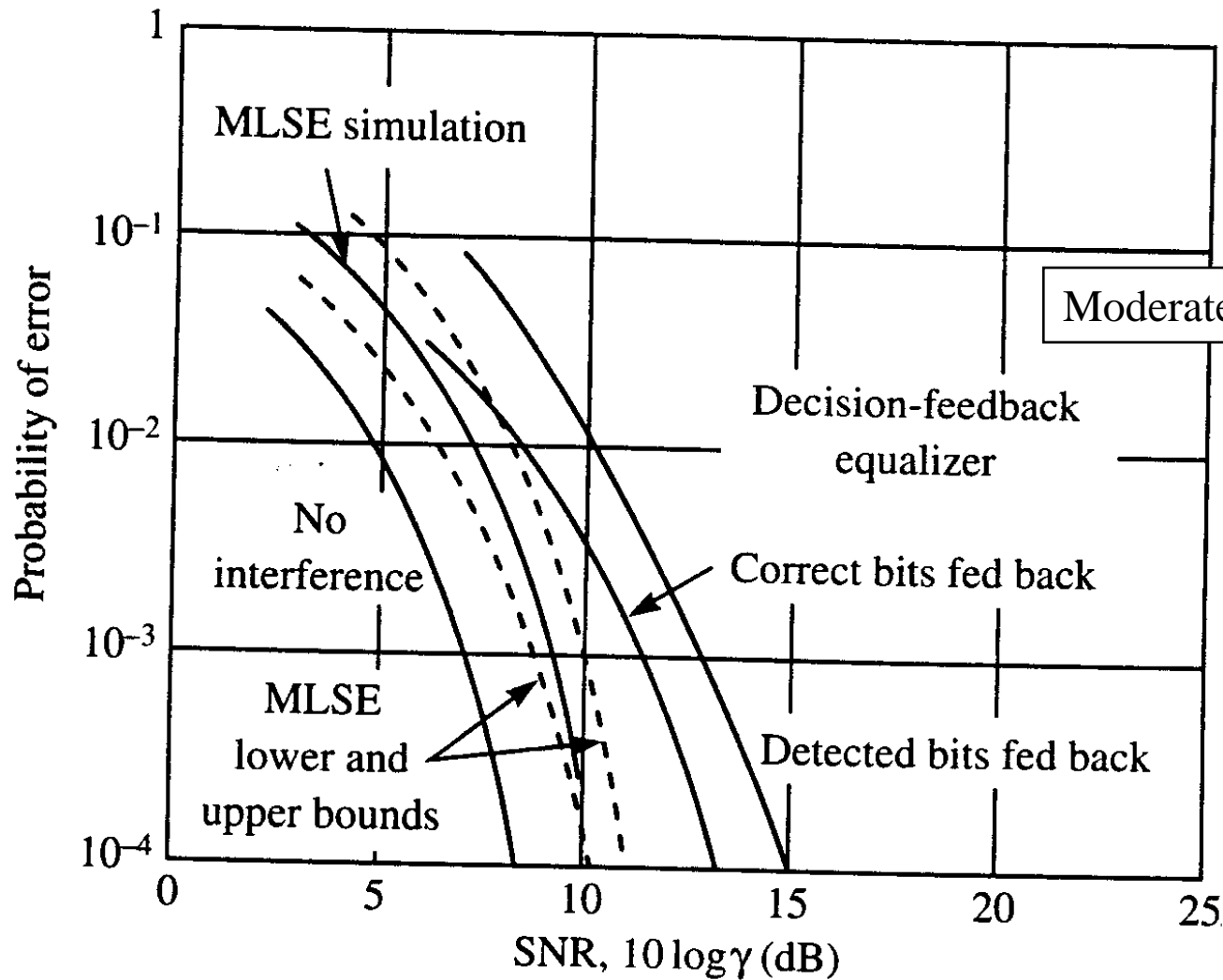
# Performance of different Equalizers



Moderate  
Frequency  
Selectivity

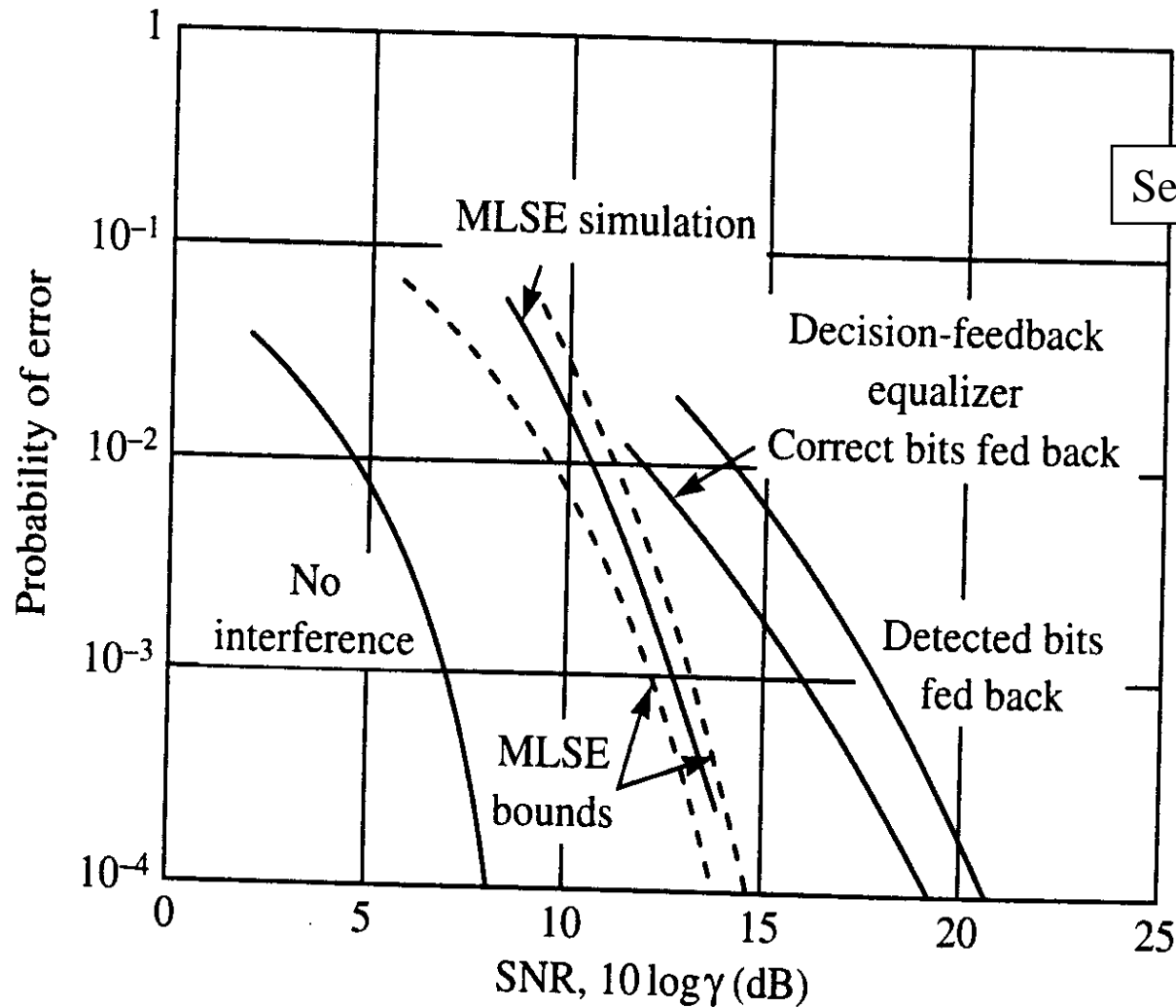
DFE can provide  
substantial gains

# Performance of Decision Feedback and MLSE Structures



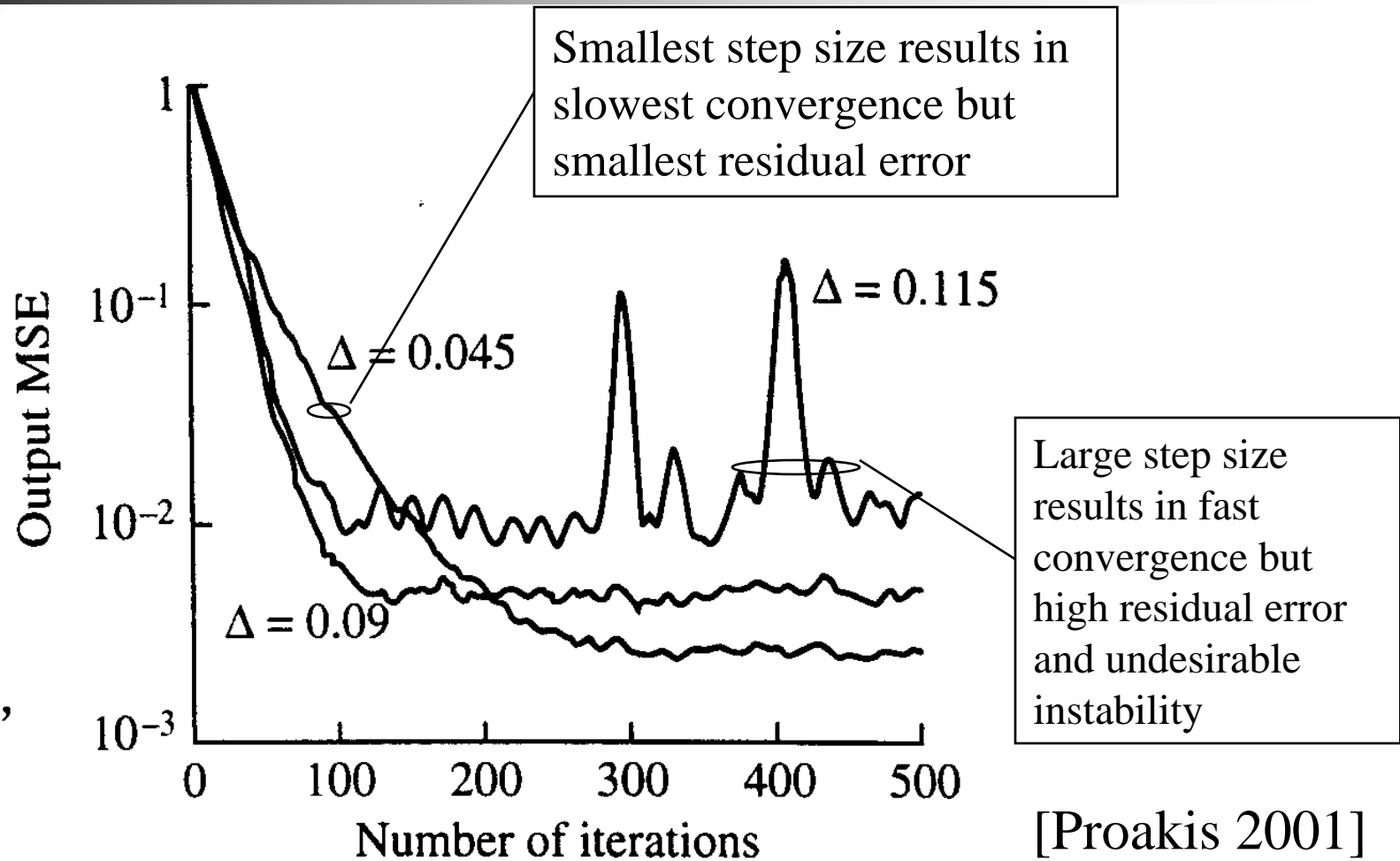
[Proakis 2001]

# Performance of Decision Feedback and MLSE Structures

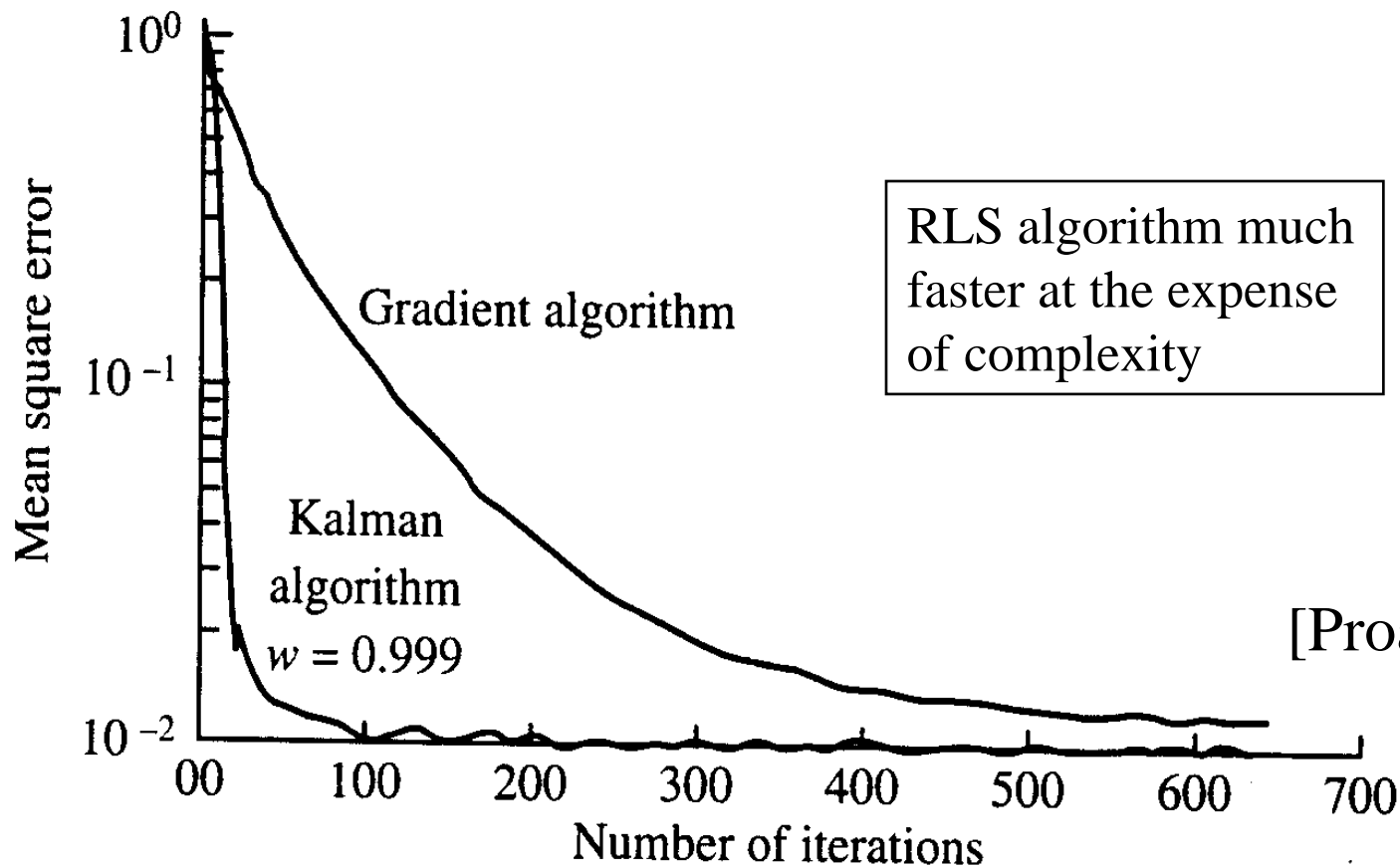


[Proakis 2001]

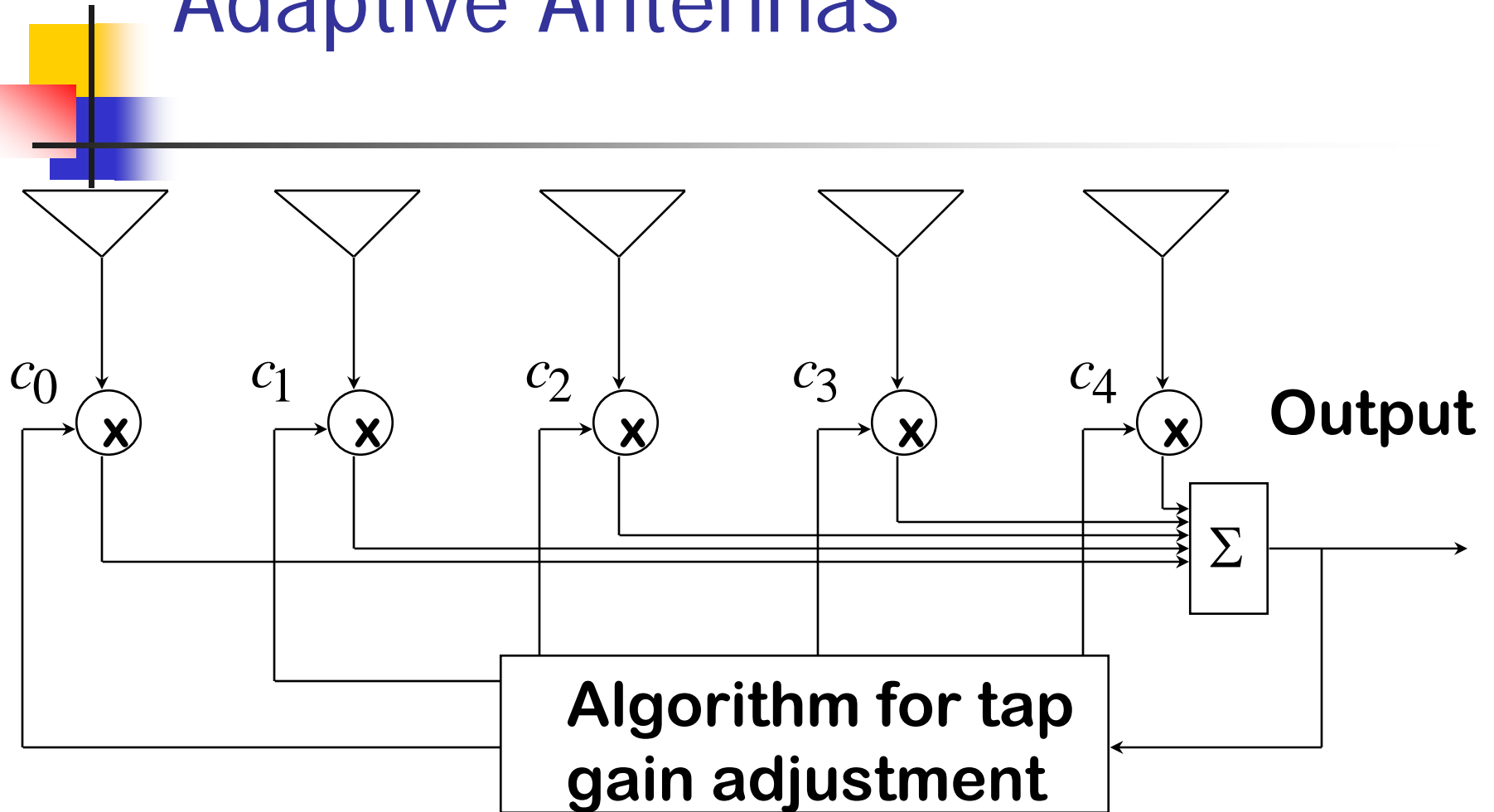
# Effect of Step Size on LMS Algorithm



# Convergence Rate of LMS (Gradient) and RLS (Kalman) Algorithms



# Adaptive Antennas





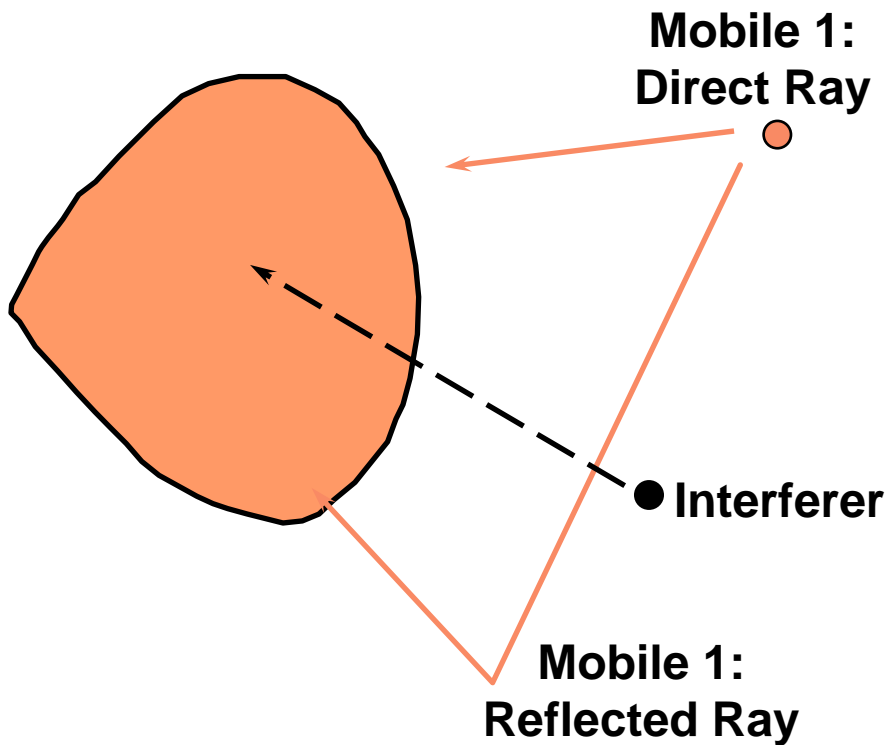
# Analogy Between Adaptive Antennas and Equalizers

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- Antenna arrays are formed by combining the signals from a number of antenna elements (frequently spaced one half wavelength apart).
- Coefficients which weight coefficients form a “spatial filter”
- Goal is usually to minimize MSE in resulting signal
- Adaptive algorithms such as LMS or RLS can be used to steer the resulting antenna beam
- Interference is analogous to ISI. We are now interested in the correlation across antennas rather than in time.

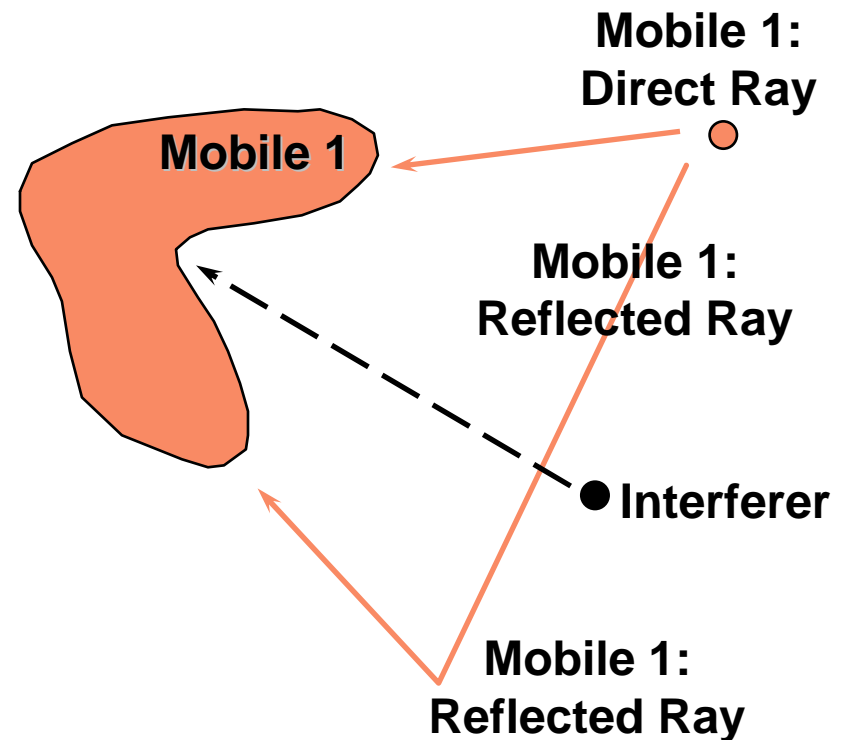
# Adaptive Antenna Illustration

## Conventional Receiver



Antenna spatial gain patterns are fixed.

## Adaptive Antenna



Adaptively "Weight" and Combine Multiple Antenna Elements To Optimize Performance

# Adaptive Antenna Array Beamforming

Multiple patterns can be created at base station for multiple users.

- 4 signals positioned in a 120° sector.
- 6 element linear array ( $\lambda/2$  spacing).
- Optimal adaptive beamforming (MMSE) for each user.

